

Dynamic behavior of more complicated processes

DINAMICA E CONTROLLO DEI PROCESSI CHIMICI

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ANDREA MIO, PhD

MolBNL @ DIA-UniT; www.molbnl.it
amio@units.it - 0405583440

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A GENERIC DYNAMIC RESPONSE

- The intrinsic dynamic characteristics of a system are determined by the denominator polynomial of the TF, also known as **characteristic polynomial**. For example:

$$G(s) = \frac{Y(s)}{U(s)} = \frac{K}{s(\tau s + 1)(\tau^2 s^2 + 2\zeta\tau s + 1)} \quad ; \quad 0 \leq \zeta < 1$$

$$Y(s) = G(s)U(s) = \frac{A}{s} + \frac{B}{\tau s + 1} + \frac{C}{\tau^2 s^2 + 2\zeta\tau s + 1} + (\text{input related terms})$$

- Independently of the input**, the response of this generic example will contain the following functions of time:

- ▷ A constant (i.e., time independent) term, resulting from the **s factor**
 - ▷ An exponential term of the e^{-t/τ_1} type, resulting from the **$(\tau_1 s + 1)$ factor**
 - ▷ $e^{-\zeta t/\tau_2} \sin\left(\frac{\sqrt{1-\zeta^2}}{\tau_2} t\right)$
 - ▷ $e^{-\zeta t/\tau_2} \cos\left(\frac{\sqrt{1-\zeta^2}}{\tau_2} t\right)$
- } terms resulting from the **$(\tau^2 s^2 + 2\zeta\tau s + 1)$ factor**

These 4 dynamic terms are «**weighted**» through constants A, B and C (whose values depend on τ_1 , τ_2 and ζ)

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NATURAL MODES AND POLES

$$G(s) = \frac{A}{\tau_1 s + 1} + \frac{B}{\tau_2 s + 1} + \frac{C}{\tau_3 s + 1} + \frac{D}{\tau_4 s + 1} + \dots = G_1(s) + G_2(s) + G_3(s) + G_4(s) + \dots \rightarrow$$

$$\rightarrow Y(s) = G(s) \cdot U(s) = G_1(s)U(s) + G_2(s)U(s) + G_3(s)U(s) + G_4(s)U(s) + \dots$$

- However complex $G(s)$ might be, it may be thought as the superposition (combination) of simpler dynamic modes
- The τ_i 's may be **real** or **complex**, with positive or negative real parts
- Each dynamic mode contributes to the overall response through an **addendum** of $G(s)$
 - The overall response is the weighted sum of the natural modes, each of which does not depend on the input
- Despite the shape of the overall response will depend on the f.f. $U(s)$, the intrinsic dynamics (which depends on natural modes only) is **independent** from $U(s)$
- **Notice:** If **one natural mode** (pole) drive one subsystem to instability, the **overall response will be unstable!**

(Barolo, 2019)

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NATURAL MODES AND POLES

/cont'd

$p = -\frac{1}{\tau}$
 $\frac{1}{\tau} e^{-t/\tau}$
 $-p e^{+t \cdot p}$

$$G(s) = \frac{Y(s)}{U(s)} = \frac{K}{s(\tau s + 1)(\tau^2 s^2 + 2\zeta\tau s + 1)}$$

Additional characteristics of $Y(s)$ response will be determined by the specific **forcing function** perturbing the system

- Anyway, the **intrinsic dynamics** of the response is determined by the **process**, not by the f.f.
 - The intrinsic characteristics are called **natural modes** o «**modi naturali**» o «**risposte naturali**»
 - From a practical p.o.v., the **natural modes** are determined by **the roots p_i** of the characteristic polynomial
 - In this case: $p_1 = 0$; $p_2 = -\frac{1}{\tau_1}$; $p_3 = -\frac{\zeta}{\tau_2} + j\frac{\sqrt{1-\zeta^2}}{\tau_2}$; $p_4 = -\frac{\zeta}{\tau_2} - j\frac{\sqrt{1-\zeta^2}}{\tau_2}$
 - In the time domain, each mode corresponds to a term of the $[-p_i e^{t \cdot p_i}]$ type $e^{(a+jb)t} = e^{at} \cdot e^{jbt}$
- The **roots** of the characteristic polynomial are called **poles** of the system
 - The natural modes are «weighted» one another depending on the values of parameters τ_i, ζ, \dots
- In general, the response is «**dominated**» by the **slowest mode** (the one with $\uparrow \tau_i$)
 - The relevant $e^{t \cdot p_i}$ decays slower

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POLES IN THE COMPLEX PLANE

$$G(s) = \frac{K}{s(\tau s + 1)(\tau^2 s^2 + 2\zeta\tau s + 1)} \quad ; 0 \leq \zeta < 1$$

- The **integrator** (p_1) is located on the **plane origin**
- Any time a **complex solution** occurs, its **conjugate** occurs as well ($p_{3,4}$)
- In the time domain, the **poles** p_i become modes of the $[-p_i e^{t \cdot p_i}]$ type

$$p_{3,4} = -\frac{\zeta}{\tau_2} \pm j \frac{\sqrt{1-\zeta^2}}{\tau_2}$$

- The **speed** of response corresponding to a given mode increases as the pole moves **further away** from the imaginary axis
- Poles with **non-null imaginary part** ($p_{3,4}$) indicate the existence of **oscillations** in the response (sine and cosine terms are present)
- Poles with **positive real part** indicate a response that does not settle to a steady state (**unstable** systems)

(Seborg et al., 2011)

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POLES IN THE COMPLEX PLANE /cont'd

Zone of system instability: $\text{Re}(p_i) > 0$
Right-half (RHP) poles

Zone of system stability: $\text{Re}(p_i) < 0$
Left-half (LHP) poles

Oscillating response: $\text{Im}(p_i) \neq 0$
Complex and conjugated poles
(on right or left half planes)

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SYSTEM WITH NUMERATOR DYNAMICS

- FO systems: $\tau \frac{dy(t)}{dt} + y(t) = Ku(t) \rightarrow G(s) = \frac{K}{\tau s + 1} \quad (\tau > 0)$
- Consider the system: $\tau_1 \frac{dy}{dt} + y = K \left(u + \tau_a \frac{du}{dt} \right)$
 - ❖ y,u deviation variables
 - ❖ System initially at s.s.
 - **Lead-lag element** o **elemento «anticipo-ritardo»**
 - It follows: $G(s) = K \frac{\tau_a s + 1}{\tau_1 s + 1}$
 - The numerator has its own dynamics!
 - A polynomial in s appears at the numerator
- Consider a different system: $\tau_1 \frac{dy}{dt} + y = K \left(u + \frac{1}{\tau_a} \int_0^t u(t^*) dt^* \right)$
 - It follows: $G(s) = K \frac{\tau_a s + 1}{\tau_a s (\tau_1 s + 1)}$
 - The numerator has its own dynamics!
 - There is an integrator term (pole at origin)

The overall dynamics of the system is determined not only by the denominator roots (poles), but also by the numerator roots (zeroes)

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TRANSFER FUNCTIONS OF DYNAMIC SYSTEMS

Pole/zero form

$$G(s) = \frac{Y(s)}{U(s)} = \frac{b_m s^m + b_{m-1} s^{m-1} + \dots + b_0}{a_n s^n + a_{n-1} s^{n-1} + \dots + a_0}$$

$m \leq n$

Gain/time constant form

$$G(s) = \frac{b_m (s - z_1)(s - z_2) \dots (s - z_m)}{a_n (s - p_1)(s - p_2) \dots (s - p_n)}$$

$$G(s) = K \frac{(\tau_a s + 1)(\tau_b s + 1) \dots}{(\tau_1 s + 1)(\tau_2 s + 1) \dots}$$

Gain $K = b_0/a_0 = G(0)$ is highlighted

- It follows that: $z_1 = -\frac{1}{\tau_a}; z_2 = -\frac{1}{\tau_b}; \dots \quad p_1 = -\frac{1}{\tau_1}; p_2 = -\frac{1}{\tau_2}; \dots$
- The presence of zeroes does **not** influence the presence (and location) of the poles
 - Unless an exact «pole-zero cancellation» occurs
- However, the zeroes have a **strong effect** on how the natural modes of the system are «weighted» one another
 - They affect the TF factorization coefficients
- **Dominant time constant:** the **largest** one (i.e., the slowest natural mode)

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STEP RESPONSE OF FO SYSTEMS WITH NUMERATOR DYNAMICS

$$u(t) = \begin{cases} 0 & t < 0 \\ M & t \geq 0 \end{cases}$$

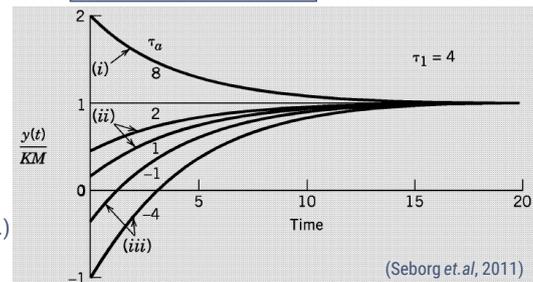
$$G(s) = K \frac{\tau_a s + 1}{\tau_1 s + 1}; \quad Y(s) = G(s)U(s); \quad U(s) = \frac{M}{s} \rightarrow Y(s) = KM \frac{\tau_a s + 1}{s(\tau_1 s + 1)}$$

$$y|_{t=0^+} = KM \left(\frac{\tau_a}{\tau_1}\right) \\ y|_{t=\infty} = KM$$

■ Upon inverse transform: $y(t) = KM \left[1 - \left(1 - \frac{\tau_a}{\tau_1}\right) e^{-t/\tau_1} \right]$

■ The response depends on the ratio $\frac{\tau_a}{\tau_1}$

- ▷ As $t \rightarrow \infty$ the output settles to **KM**, regardless of the value of the ratio
- ▷ At $t = 0^+$ the response always shows a «jump» (with respect to the initial s.s.)
 - ❖ Whether the jump is «above» or «below» the final value KM depends on the relative values of τ_a and τ_1
- ▷ If $\tau_a = 0$ a typical FO response is obtained
- ▷ If $\tau_a = \tau_1 > 0$ the response is that of a pure gain (pole-zero cancellation)
- ▷ If $\tau_a < 0$ (positive zero), the response starts below zero (i.e., below the initial steady state)



(Seborg et al, 2011)

Step response of a lead/lag element
 τ_1 : lag time; τ_a : lead time

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PHYSICAL REALIZABILITY OF A DYNAMIC SYSTEM

- Generally, a «jump» at $t=0$ (for a step f.f.) can occur **only** if the numerator $N(s)$ and the denominator $D(s)$ of $G(s)$ have **dynamics of the same order**
 - ▷ See initial value theorem, for example on: $Y(s) = K \frac{\tau_a s + 1}{(\tau_1 s + 1)} \cdot \frac{M}{s}$
- Real physical systems have $D(s)$ with dynamics of higher order with respect to that of $N(s)$
 - ▷ i.e., they have an intrinsic inertia, which prevents them from instantaneously responding to an input change
 - ▷ $m < n$: physical realizability condition
- A lead/lag response does not exist physically in industrial processes, but it can be obtained as the output of a control system
 - ▷ It is possible to build a lead-lag element digitally
 - ▷ In order to change the system dynamics, it introduces a «jump» in a manipulated input

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STEP RESPONSE OF SO SYSTEMS WITH NUMERATOR DYNAMICS

$$u(t) = \begin{cases} 0 & t < 0 \\ M & t \geq 0 \end{cases}$$

$$G(s) = K \frac{\tau_a s + 1}{(\tau_1 s + 1)(\tau_2 s + 1)}; \quad Y(s) = G(s)U(s); \quad U(s) = \frac{M}{s} \quad \rightarrow \quad Y(s) = KM \frac{\tau_a s + 1}{s(\tau_1 s + 1)(\tau_2 s + 1)}$$

$(\tau_1 > 0; \tau_2 > 0)$

■ Upon inverse transform:
$$y(t) = KM \left(1 + \frac{\tau_a - \tau_1}{\tau_1 - \tau_2} e^{-t/\tau_1} + \frac{\tau_a - \tau_2}{\tau_2 - \tau_1} e^{-t/\tau_2} \right)$$

SO response with numerator dynamics

(Seborg et al, 2011)

SO without/with numerator dynamics

(Barolo, 2019)

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STEP RESPONSE OF SO SYSTEMS WITH NUMERATOR DYNAMICS /cont'd

$$u(t) = \begin{cases} 0 & t < 0 \\ M & t \geq 0 \end{cases}$$

$$G(s) = K \frac{\tau_a s + 1}{(\tau_1 s + 1)(\tau_2 s + 1)}$$

- Compared to a SO system, adding a zero:
 - ▷ Does **not** change either the **final** value, nor the **initial** one
 - ▷ Does change the way the natural modes are «**weighted**» to provide the overall response
- For $\tau_a > 0$ (**negative zero**): the initial response is **accelerated** with respect to pure SO ($\tau_a = 0$)
 - ▷ Generally, this acceleration is independent from the denominator order
- For $\tau_a > \max(\tau_1; \tau_2)$: the initial response is accelerated up to the point that an **overshoot** occurs
 - ▷ The overall response is **slower**, even though the initial one is faster
- For $\tau_a < 0$ (**positive zero**): an **inverse** response occurs (with undershoot)

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INVERSE RESPONSE SYSTEMS

- The **initial** response to a step input change points to a direction that is **opposite** to the final steady-state one
- An inverse response arises from two different dynamic effects:
 - ▷ Are **parallel**
 - ▷ Are in **opposition**
 - ▷ Operate on **different time scales**

The combination of parallel and opposition effects give rise to **competing** effects

 - ❖ A **fast** dynamic mode, with small steady-state magnitude, is responsible for the **initial response** (the inverse behavior)
 - ❖ A **slow** dynamic mode, with dominant steady-state magnitude, is responsible for the **final steady state** behavior

A **necessary** condition for the occurrence of inverse response is the presence of a **positive zero** in the system TF

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INVERSE RESPONSE SYSTEMS /cont'd

$\tau_1 > \tau_2 > 0$

⇒

- More on necessary conditions:
 - ▷ The **gains** must have **opposite** signs
 - ▷ The «fast» process, which is the one with smaller time constant) must have **smaller** gain (in absolute value)
- The **overall gain** $K = K_1 + K_2$ has the same **sign** as the one with **dominant effect**, which is the one slower with the higher value of τ

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WHERE CAN WE FIND INVERSE RESPONSE SYSTEMS?

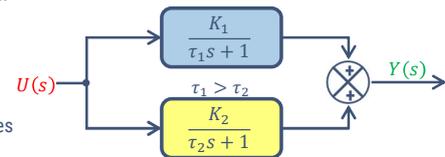
■ Open-loop inverse responses of real physical systems

- ▷ Response of the **bottom level of a distillation column** to a step increase of the reboiler steam flow

- ❖ **Long-term effect:** more liquid is evaporated from the bottom → the bottom level decreases
- ❖ **Short-term effect:** more liquid overflows from tray weirs → the bottom level increases

- ▷ Response of the **exit temperature in an exothermic catalytic tubular reactor** to a step increase of the feed temperature

- ❖ **Long-term effect:** the catalyst T increases all along the catalyst bed → the T of reactants close to the exit increases → the speed of reaction in this section increases → the product exit T increases
- ❖ **Short-term effect:** the conversion close to the reactor inlet increase before the catalyst can be heated up → the reactants concentration close to the exit decreases because of fluid transportation → less heat is released close to the exit → exit T decreases



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