

Studio dei sistemi dinamici tramite FdT

Risposta allo scalino

Introduzione

- Assegnato un **sistema dinamico LTI** descritto tramite una **Funzione di Trasferimento** (a tempo continuo oppure a tempo discreto), e ' possibile trovare dei parametri che descrivano l'evoluzione della risposta allo scalino del sistema e che siano in qualche modo legati ai parametri tipici della FdT (zeri/poli, guadagno statico ecc.)?

Tempo continuo vs tempo discreto

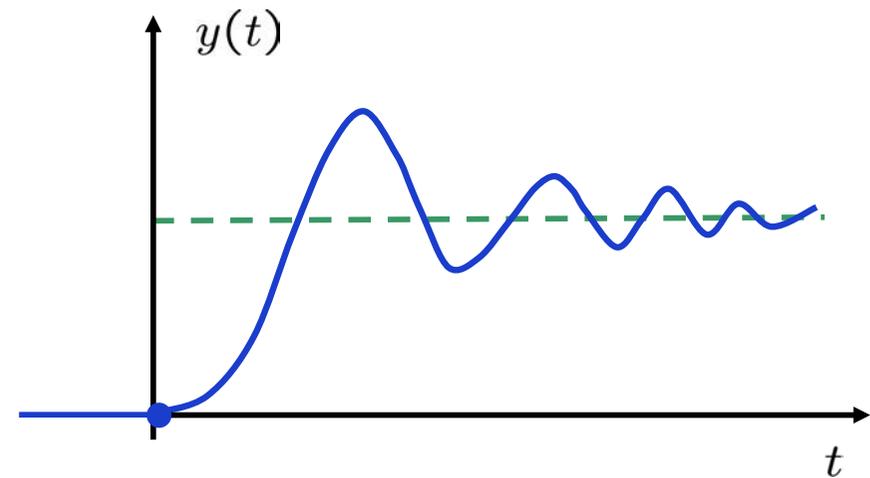
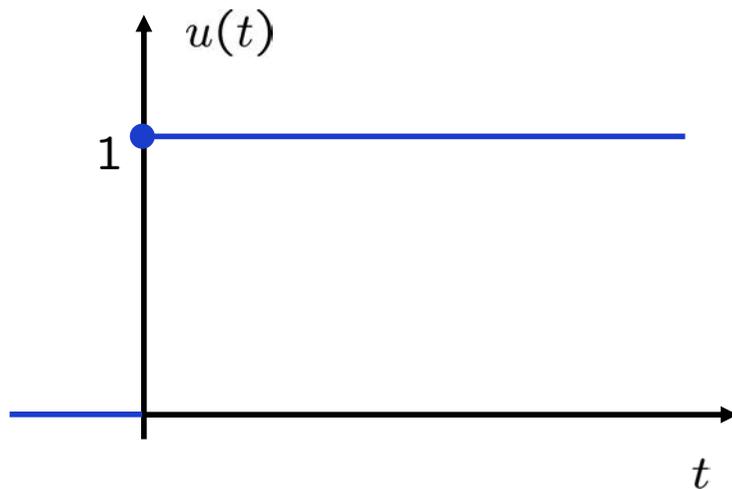
- Analizziamo **separatamente** anche stavolta il caso dei **sistemi LTI a tempo continuo** ed il caso dei **sistemi LTI a tempo discreto**.
- Infatti, come vedremo, i risultati dell'analisi svolta su sistemi a tempo continuo non sono tutti direttamente portabili al caso dei sistemi a tempo discreto e viceversa esistono dei comportamenti particolari che sono peculiarità dei sistemi a tempo discreto.

Risposta allo scalino di sistemi LTI a tempo continuo

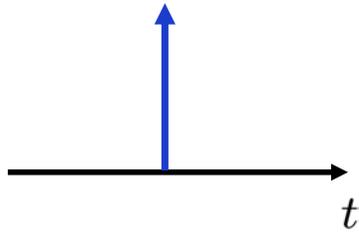
- Risposta allo scalino

$$x(0) = 0$$

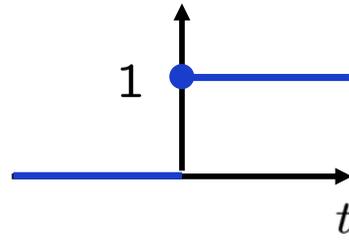
$$u(t) = 1(t)$$



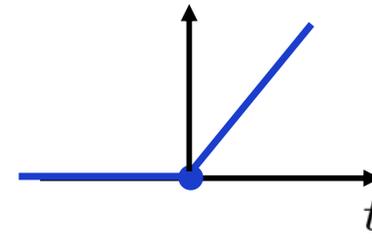
In sistemi asint. stabili descrive la transizione da un equilibrio ad un altro



$$\delta(t)$$



$$1(t)$$



$$t \cdot 1(t)$$

$$\frac{d}{dt} 1(t)$$

$$1$$

$$\int_0^t 1(\tau) d\tau$$

$$\frac{1}{s^2}$$

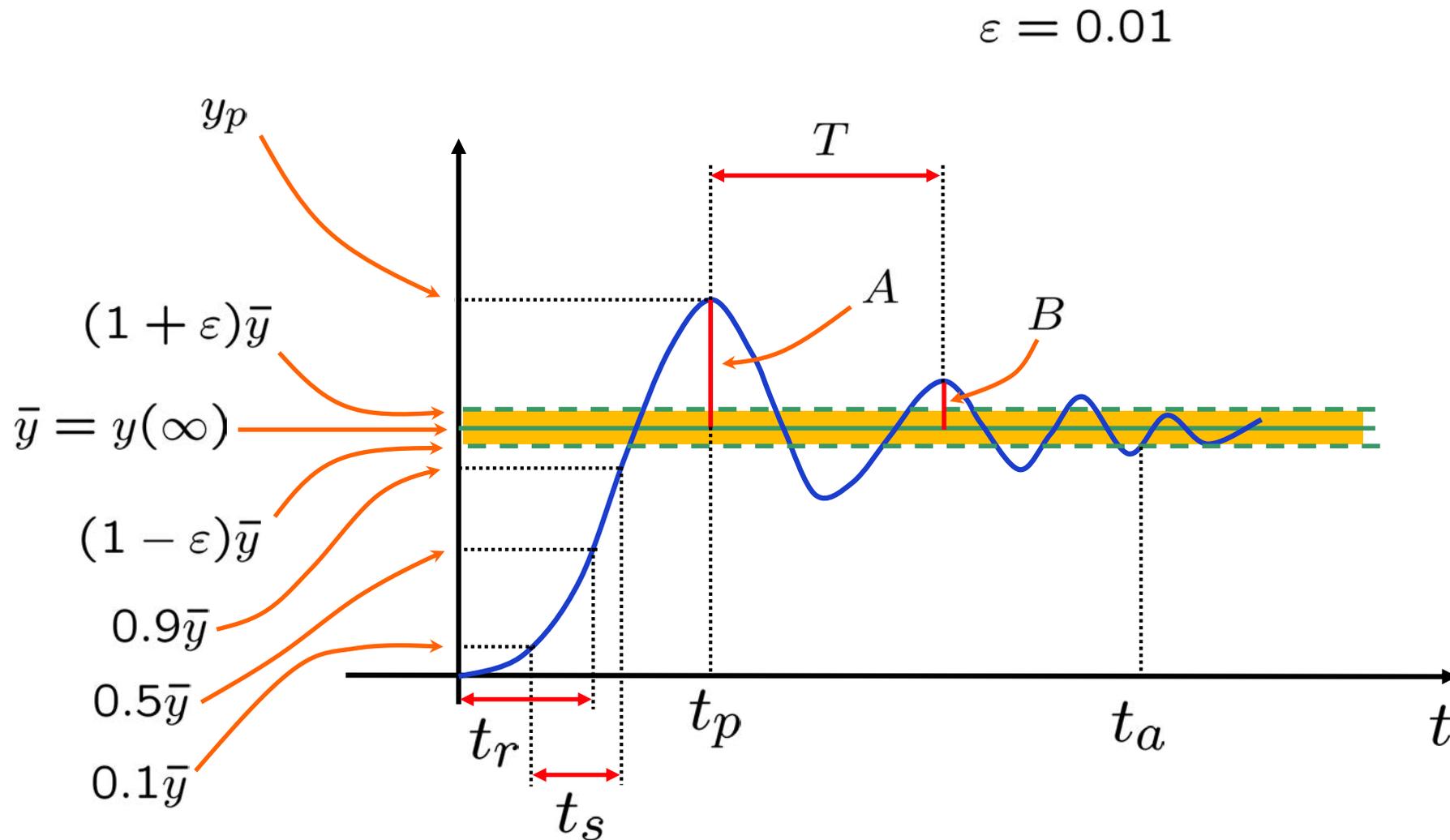
Risposta all'impulso = $\frac{d}{dt}$ (risposta allo scalino)

Risposta alla rampa = \int_0^t (risposta allo scalino) $d\tau$

- Risposta allo scalino: parametri caratteristici

- $\bar{y} = y(\infty)$ Valore di regime
- t_a Tempo di assestamento
- t_s Tempo di salita
- t_r Tempo di ritardo
- t_p Tempo di picco
- y_p Valore di picco
- $A = y_p - y(\infty)$ Max sovraelongazione
- $\Delta = A/y(\infty)$ Max sovraelongazione relativa
- T "Periodo" delle oscillazioni
- B/A Fattore di smorzamento

- Risposta allo scalino: parametri caratteristici



- Risposta allo scalino: I ordine

A)
$$G(s) = \frac{\mu}{1 + s\tau}$$

Strettamente proprio

B)
$$G(s) = \frac{\mu(1 + sT)}{1 + s\tau}$$

Non strettamente proprio

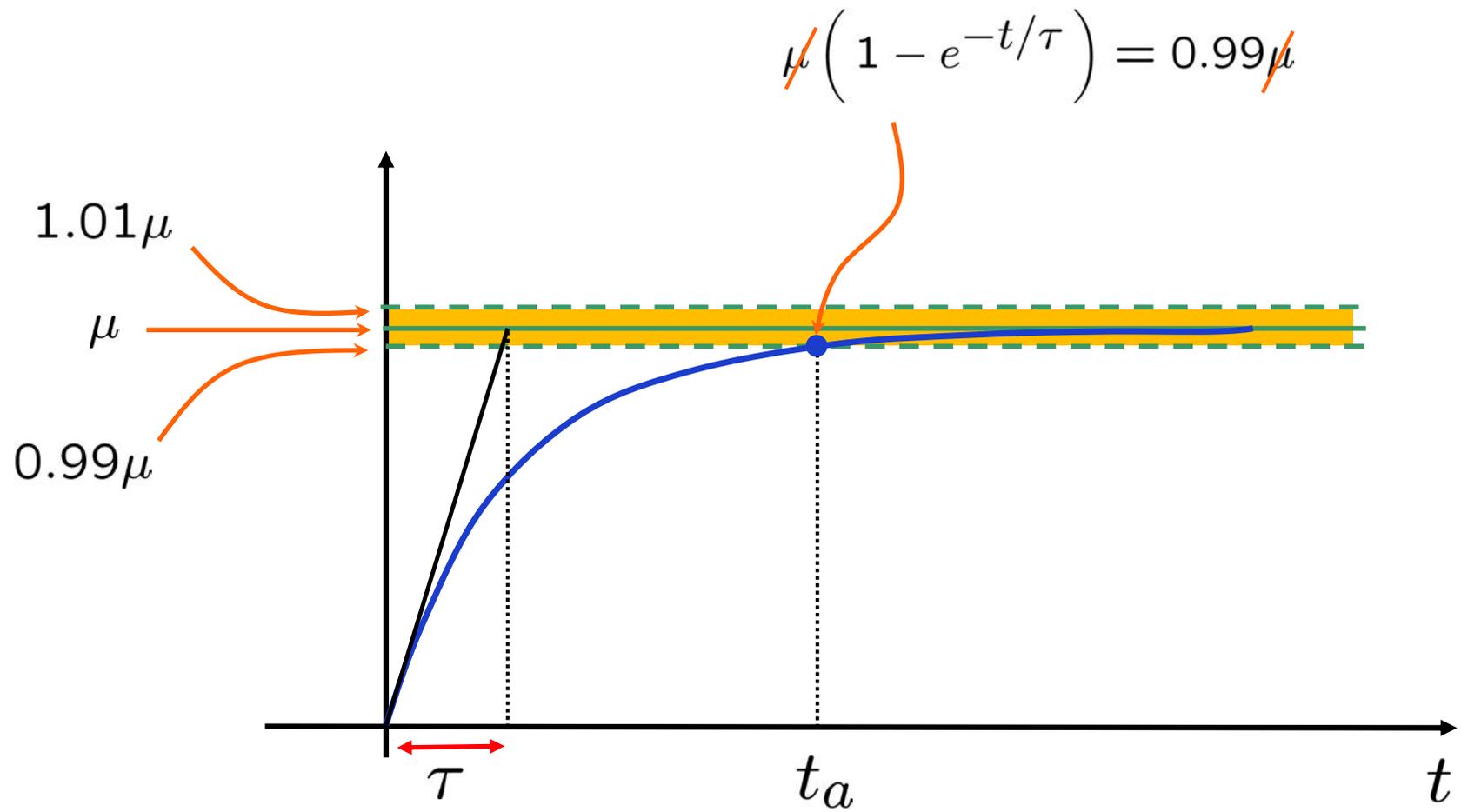
$$\tau > 0$$

$$\mu > 0$$

(As. Stabile)

A)

$$\begin{aligned}y(t) &= \mathcal{L}^{-1} \left[G(s) \frac{1}{s} \right] = \mathcal{L}^{-1} \left[\frac{\mu}{s(1 + s\tau)} \right] \\&= \mathcal{L}^{-1} \left[\frac{\mu}{s} - \frac{\mu\tau}{1 + s\tau} \right] \\&= \mu \left(1 - e^{-t/\tau} \right), \quad t \geq 0\end{aligned}$$



- Calcolo tempo di assestamento

$$1 - e^{-t/\tau} = 0.99$$

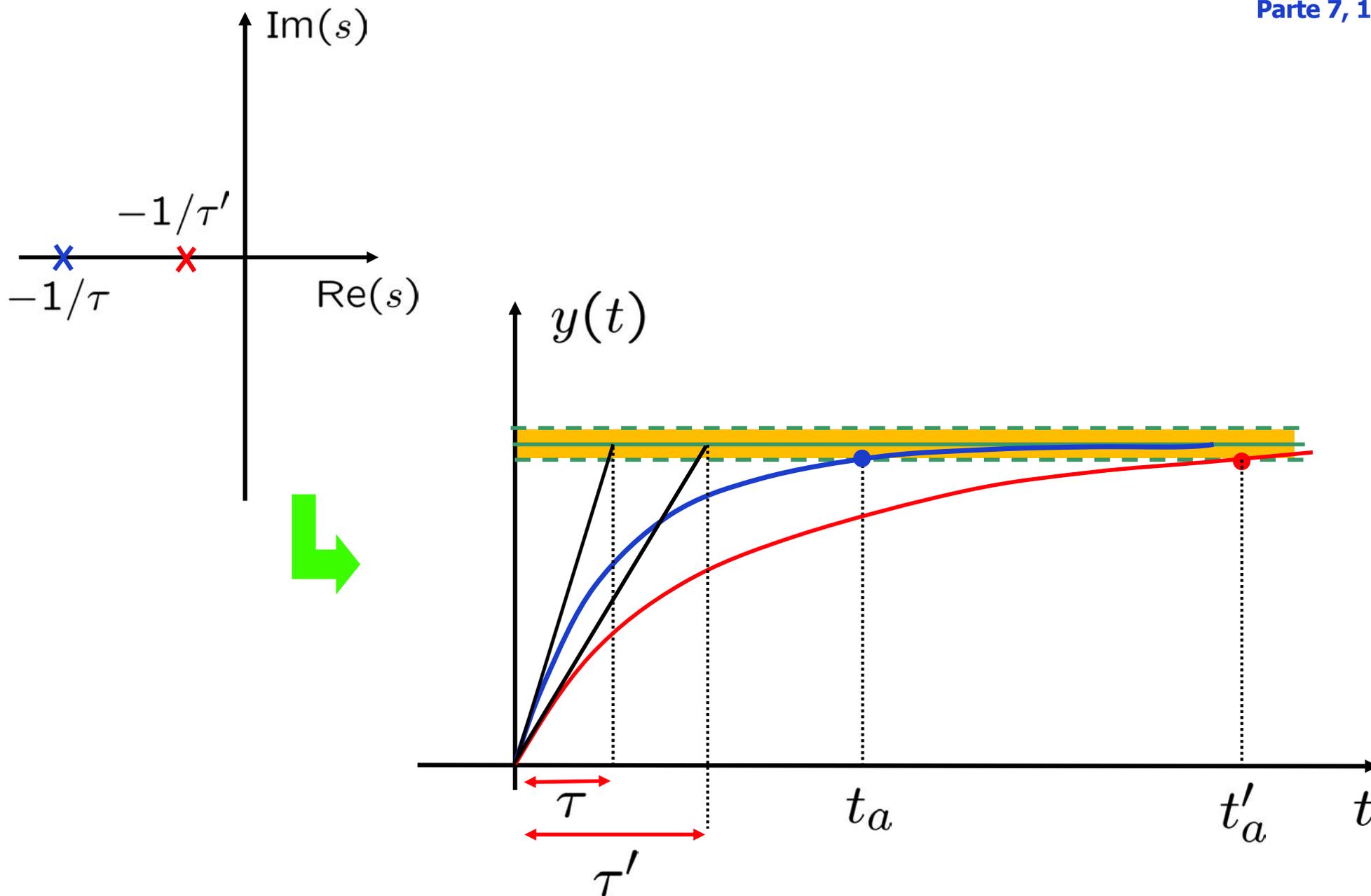

$$e^{-t/\tau} = 0.01$$


$$e^{t/\tau} = 100$$


$$t/\tau = \ln 100$$


$$t_a = \tau \ln 100 \simeq 5\tau$$

Esercizio a casa: calcolare t_s , t_r

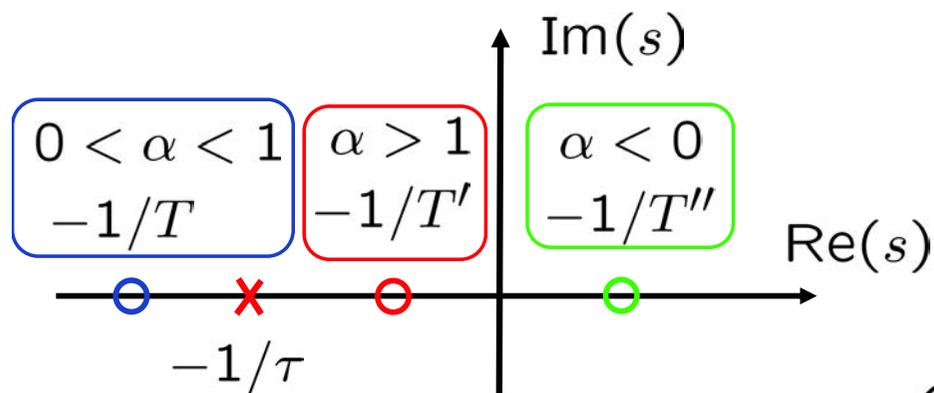


B)

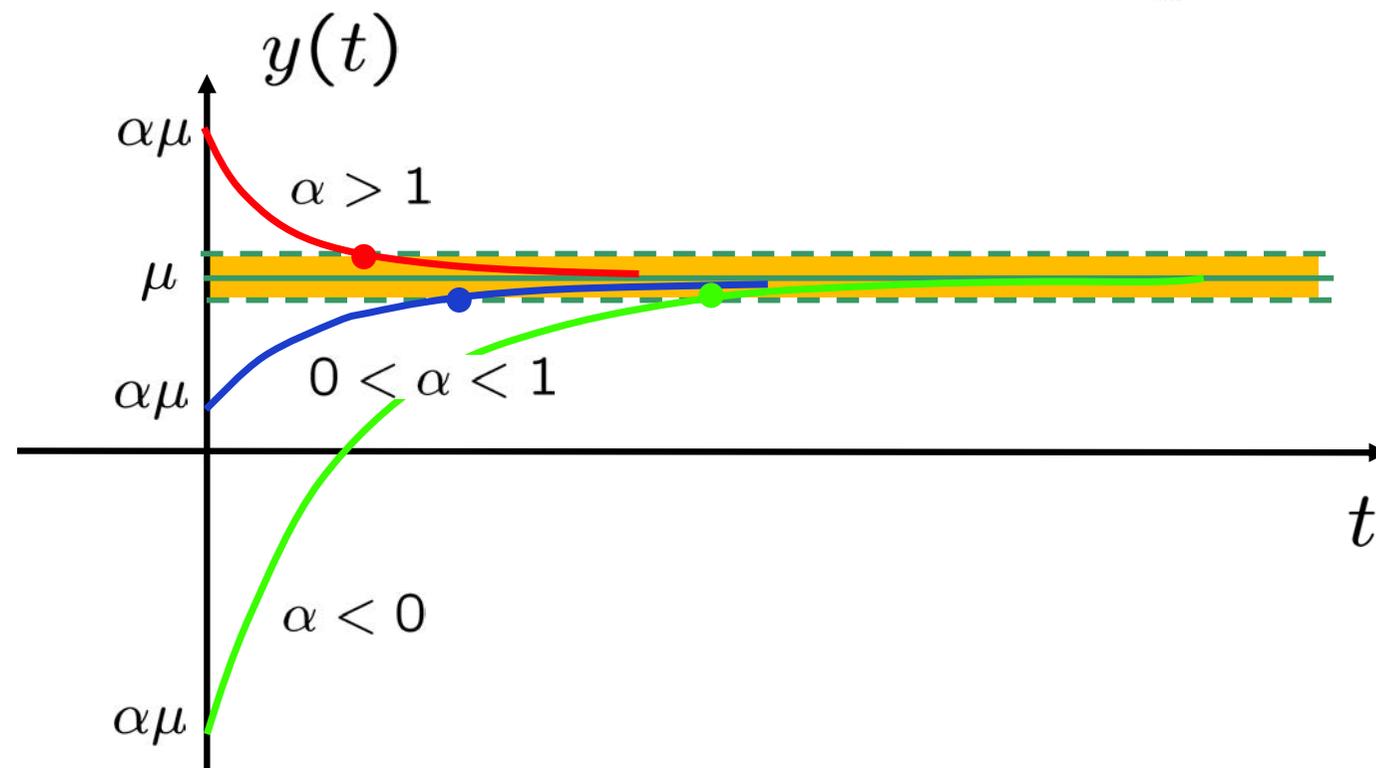
$$\begin{aligned}y(t) &= \mathcal{L}^{-1} \left[G(s) \frac{1}{s} \right] = \mathcal{L}^{-1} \left[\frac{\mu(1 + sT)}{s(1 + s\tau)} \right] \\&= \mathcal{L}^{-1} \left[\frac{\mu}{s} + \frac{\mu(T - \tau)}{1 + s\tau} \right] \\&= \mu \left(1 + \frac{T - \tau}{\tau} e^{-t/\tau} \right), \quad t \geq 0\end{aligned}$$

N.B. $\lim_{t \rightarrow 0^+} y(t) = \mu \frac{T}{\tau} \neq 0$

Poniamo: $T = \alpha\tau$ \longrightarrow $y(t) = \mu \left[1 + (\alpha - 1)e^{-t/\tau} \right]$, $t \geq 0$



In prima approx: $t_a \simeq 5\tau$



- Risposta allo scalino: II ordine

$$\mathbf{A)} \quad G(s) = \frac{\mu}{(1 + s\tau_1)(1 + s\tau_2)}$$

Poli reali senza zeri

$$\mathbf{B)} \quad G(s) = \frac{\mu(1 + sT)}{(1 + s\tau_1)(1 + s\tau_2)}$$

Poli reali con zero

$$\mathbf{C)} \quad G(s) = \frac{\varrho}{(s + \sigma + j\omega)(s + \sigma - j\omega)}$$

Poli complessi senza zeri

$$\mathbf{D)} \quad G(s) = \frac{\varrho(1 + sT)}{(s + \sigma + j\omega)(s + \sigma - j\omega)}$$

Poli complessi con zero

$$\mathbf{A)} \quad G(s) = \frac{\mu}{(1 + s\tau_1)(1 + s\tau_2)}$$

$$\tau_1 \neq \tau_2$$

$$\tau_1 > 0$$

$$\tau_2 > 0$$

} Asintotica stabilita`

Per esempio poniamo: $\tau_1 > \tau_2$

$$\mu > 0$$

$$y(t) = \mathcal{L}^{-1} \left[G(s) \frac{1}{s} \right] = \mathcal{L}^{-1} \left[\frac{\mu}{s(1 + s\tau_1)(1 + s\tau_2)} \right]$$

$$= \mathcal{L}^{-1} \left[\frac{A}{s} + \frac{B}{1 + s\tau_1} + \frac{C}{1 + s\tau_2} \right]$$

$$A = \frac{\mu}{(1 + s\tau_1)(1 + s\tau_2)} \Big|_{s=0} = \mu$$

$$B = \frac{\mu}{s(1 + s\tau_2)} \Big|_{s=-1/\tau_1} = \frac{\mu}{-\frac{1}{\tau_1} \left(1 - \frac{\tau_2}{\tau_1}\right)} = \frac{\mu\tau_1^2}{\tau_2 - \tau_1}$$

$$C = \frac{\mu}{s(1 + s\tau_1)} \Big|_{s=-1/\tau_2} = \frac{\mu}{-\frac{1}{\tau_2} \left(1 - \frac{\tau_1}{\tau_2}\right)} = \frac{\mu\tau_2^2}{\tau_1 - \tau_2}$$



$$y(t) = \mathcal{L}^{-1} \left[\frac{\mu}{s} + \frac{\frac{\mu\tau_1^2}{\tau_2 - \tau_1}}{1 + s\tau_1} + \frac{\frac{\mu\tau_2^2}{\tau_1 - \tau_2}}{1 + s\tau_2} \right]$$

$$= \mu \left(1 - \frac{\tau_1}{\tau_1 - \tau_2} e^{-t/\tau_1} + \frac{\tau_2}{\tau_1 - \tau_2} e^{-t/\tau_2} \right), \quad t \geq 0$$

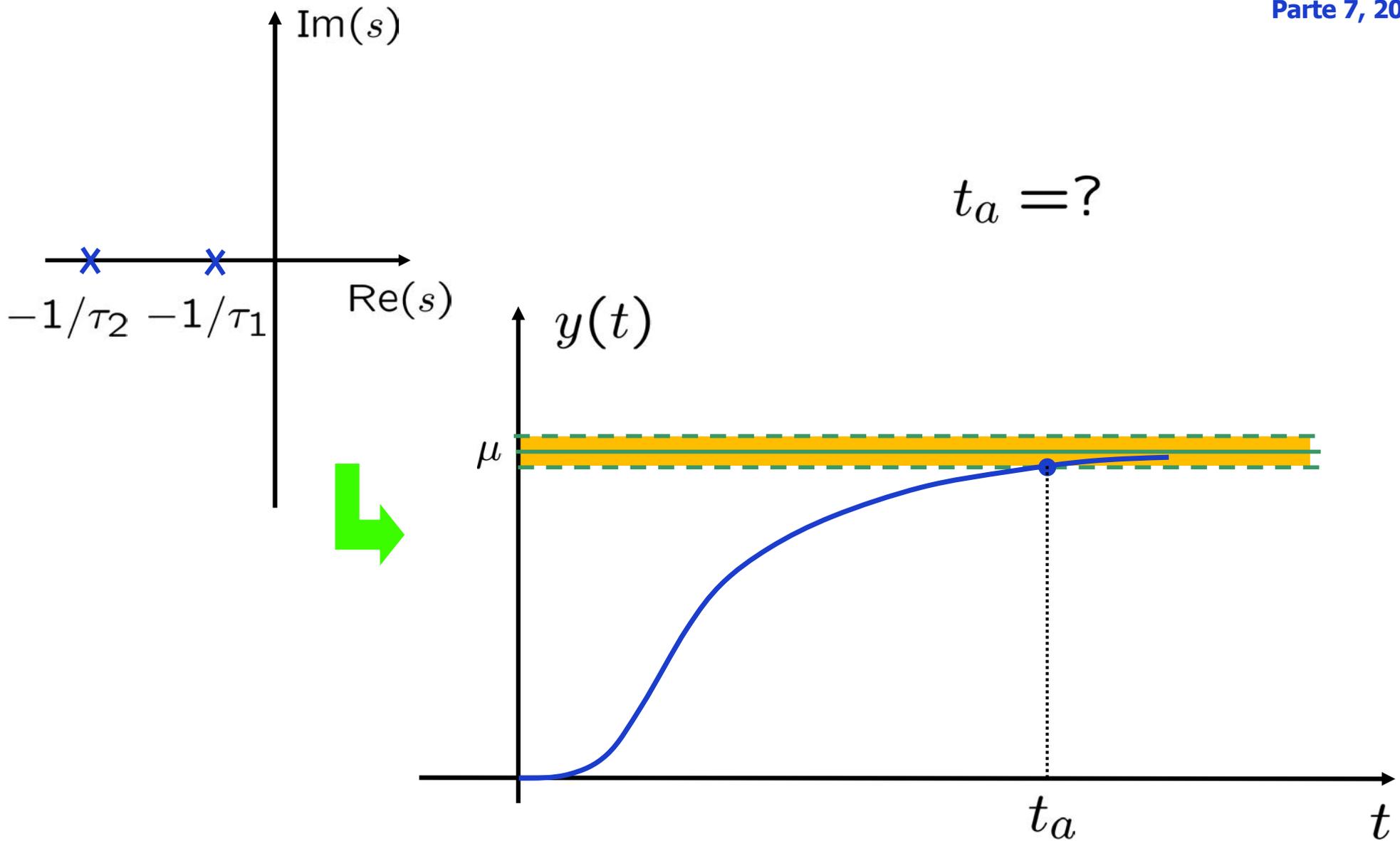
Alcune caratteristiche:

$$y(\infty) = \mu > 0$$

$$y(0) = 0$$

$$\dot{y}(0) = 0$$

$$\ddot{y}(0) = \frac{\mu}{\tau_1\tau_2} > 0$$



Curva monotona crescente, con flesso

- Se $\tau_1 \gg \tau_2$

$$\rightarrow y(t) = \mu \left(1 - \frac{\tau_1}{\tau_1 - \tau_2} e^{-t/\tau_1} + \frac{\tau_2}{\tau_1 - \tau_2} e^{-t/\tau_2} \right), \quad t \geq 0$$

$$\simeq \mu \left(1 - e^{-t/\tau_1} \right), \quad t \geq 0$$

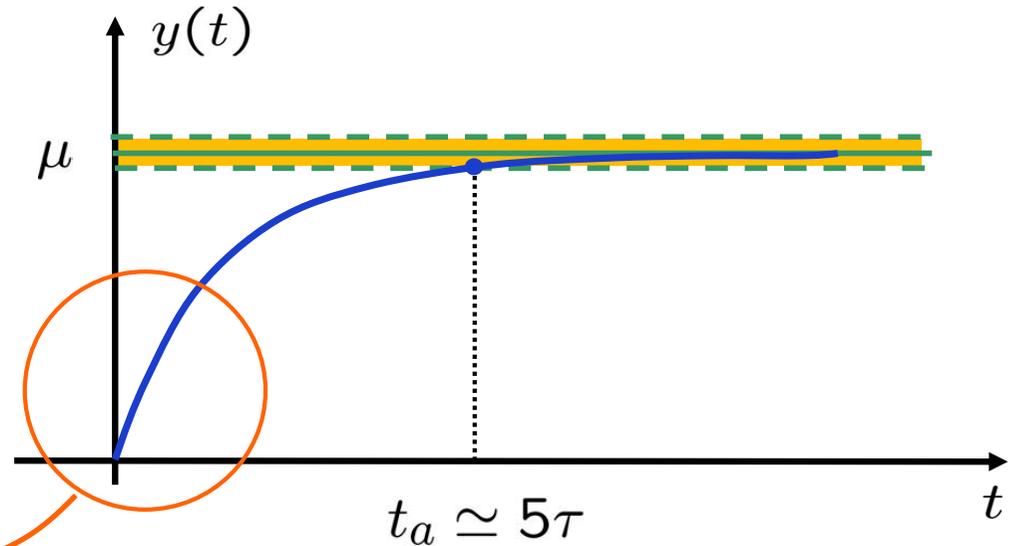


In prima approx: $t_a \simeq 5\tau_1$

In generale, in assenza di zeri, i poli più influenti sull'andamento qualitativo della risposta sono quelli vicini all'asse immaginario

- Riepilogando:

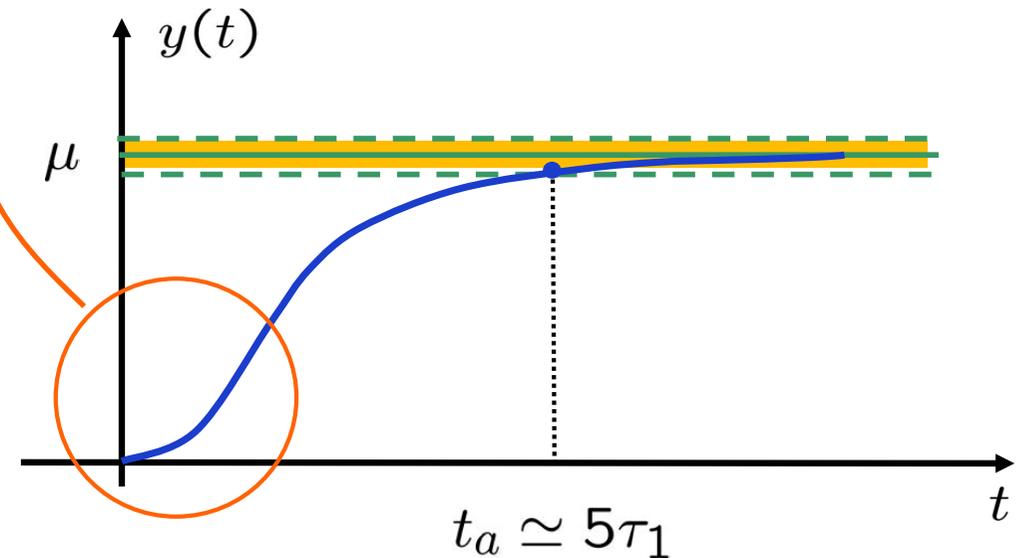
$$G(s) = \frac{\mu}{1 + s\tau}$$



A parita` di tempo di assestamento e` diverso il tempo di salita, ovvero la velocita` di risposta

$$G(s) = \frac{\mu}{(1 + s\tau_1)(1 + s\tau_2)}$$

$$\tau_1 \gg \tau_2$$



$$\mathbf{B)} \quad G(s) = \frac{\mu(1 + sT)}{(1 + s\tau_1)(1 + s\tau_2)}$$

$$\tau_1 \neq \tau_2$$

$$\tau_1 > 0$$

$$\tau_2 > 0$$

} Asintotica stabilita`

Per esempio poniamo: $\tau_1 > \tau_2$

$$\mu > 0$$

$$y(t) = \mathcal{L}^{-1} \left[G(s) \frac{1}{s} \right] = \mathcal{L}^{-1} \left[\frac{\mu(1 + sT)}{s(1 + s\tau_1)(1 + s\tau_2)} \right]$$

$$= \mathcal{L}^{-1} \left[\frac{A}{s} + \frac{B}{1 + s\tau_1} + \frac{C}{1 + s\tau_2} \right]$$

$$A = \frac{\mu(1 + sT)}{(1 + s\tau_1)(1 + s\tau_2)} \Big|_{s=0} = \mu$$

$$B = \frac{\mu(1 + sT)}{s(1 + s\tau_2)} \Big|_{s=-1/\tau_1} = \frac{\mu(1 - T/\tau_1)}{-\frac{1}{\tau_1}(1 - \frac{\tau_2}{\tau_1})} = \frac{\mu\tau_1(\tau_1 - T)}{\tau_2 - \tau_1}$$

$$C = \frac{\mu(1 + sT)}{s(1 + s\tau_1)} \Big|_{s=-1/\tau_2} = \frac{\mu(1 - T/\tau_2)}{-\frac{1}{\tau_2}(1 - \frac{\tau_1}{\tau_2})} = \frac{\mu\tau_2(\tau_2 - T)}{\tau_1 - \tau_2}$$


$$y(t) = \mu \left(1 - \frac{\tau_1 - T}{\tau_1 - \tau_2} e^{-t/\tau_1} + \frac{\tau_2 - T}{\tau_1 - \tau_2} e^{-t/\tau_2} \right), \quad t \geq 0$$

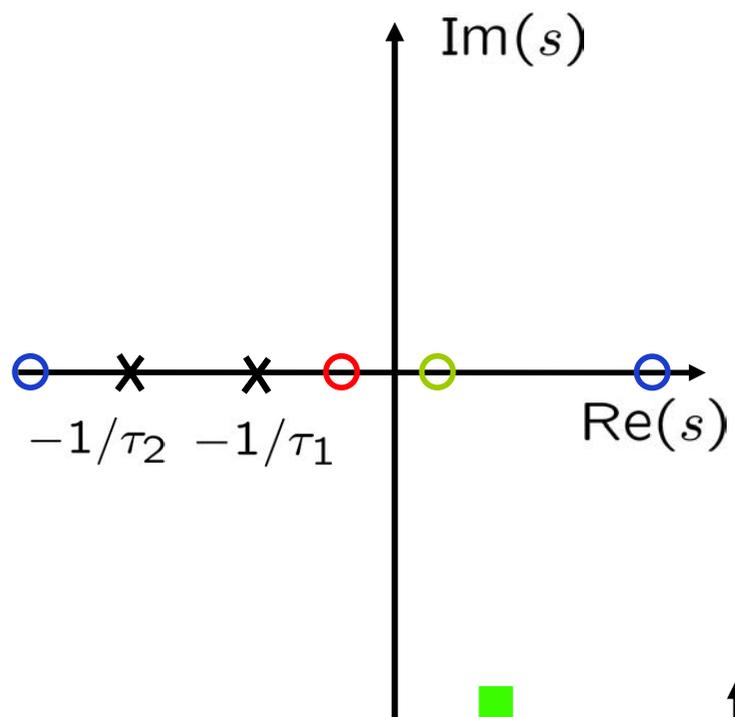
Alcune caratteristiche:

$$y(\infty) = \mu > 0$$

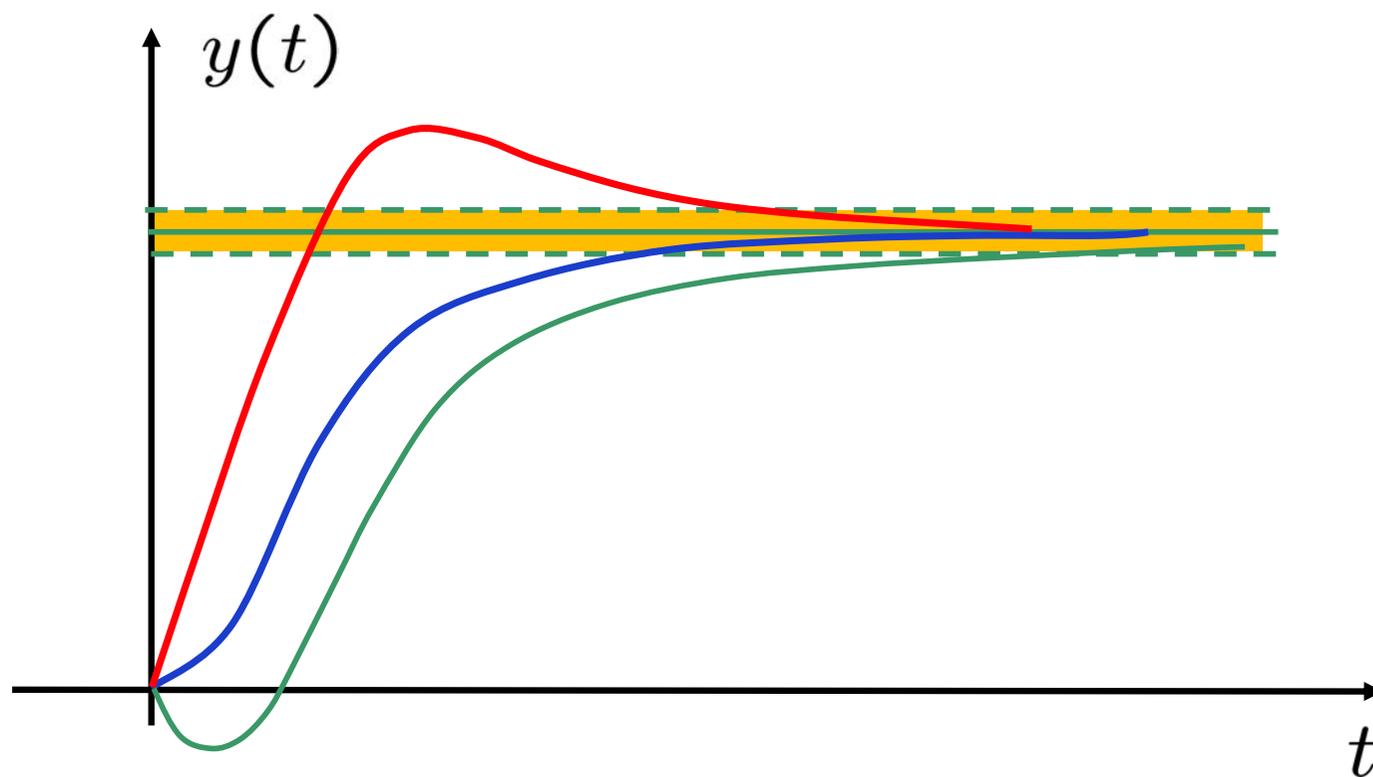
$$y(0) = 0$$

$$\dot{y}(0) = \frac{\mu T}{\tau_1 \tau_2} > 0, \quad \text{se } T > 0$$

$$< 0, \quad \text{se } T < 0$$



- zero approx irrilevante
- sovraelongazione
- sottoelongazione



$$c) \quad G(s) = \frac{\varrho}{(s + \sigma + j\omega)(s + \sigma - j\omega)}$$

$$\mu = G(0) = \frac{\varrho}{\sigma^2 + \omega^2}$$

Poli: $-\sigma \pm j\omega$

$\sigma > 0$ asintotica stabilita`

$\omega > 0$

$\varrho > 0$

$$Y(s) = \frac{G(s)}{s} = \frac{A}{s} + \frac{Bs + C}{s^2 + 2\sigma s + \sigma^2 + \omega^2}$$

$$\rightarrow As^2 + 2A\sigma s + A\sigma^2 + A\omega^2 + Bs^2 + Cs = \rho$$

$$\rightarrow \begin{cases} A + B = 0 \\ 2A\sigma + C = 0 \\ A(\sigma^2 + \omega^2) = \rho \end{cases} \rightarrow \begin{cases} A = \frac{\rho}{\sigma^2 + \omega^2} = \mu \\ B = -\mu \\ C = -2\sigma\mu \end{cases}$$

$$\rightarrow Y(s) = \mu \left[\frac{1}{s} - \frac{s + 2\sigma}{s^2 + 2\sigma s + \sigma^2 + \omega^2} \right] = \mu \left[\frac{1}{s} - \frac{s + \sigma + \sigma}{(s + \sigma)^2 + \omega^2} \right]$$

$$= \mu \left[\frac{1}{s} - \frac{s + \sigma}{(s + \sigma)^2 + \omega^2} - \frac{\sigma}{\omega} \frac{\omega}{(s + \sigma)^2 + \omega^2} \right]$$



$$y(t) = \mu \left[1 - e^{-\sigma t} \cos(\omega t) - \frac{\sigma}{\omega} e^{-\sigma t} \sin(\omega t) \right], \quad t \geq 0$$

$$= \mu \left[1 - e^{-\sigma t} \left(\cos(\omega t) + \frac{\sigma}{\omega} \sin(\omega t) \right) \right], \quad t \geq 0$$

Oscillazioni smorzate

Alcune caratteristiche:

$$y(\infty) = \mu > 0$$

$$y(0) = 0$$

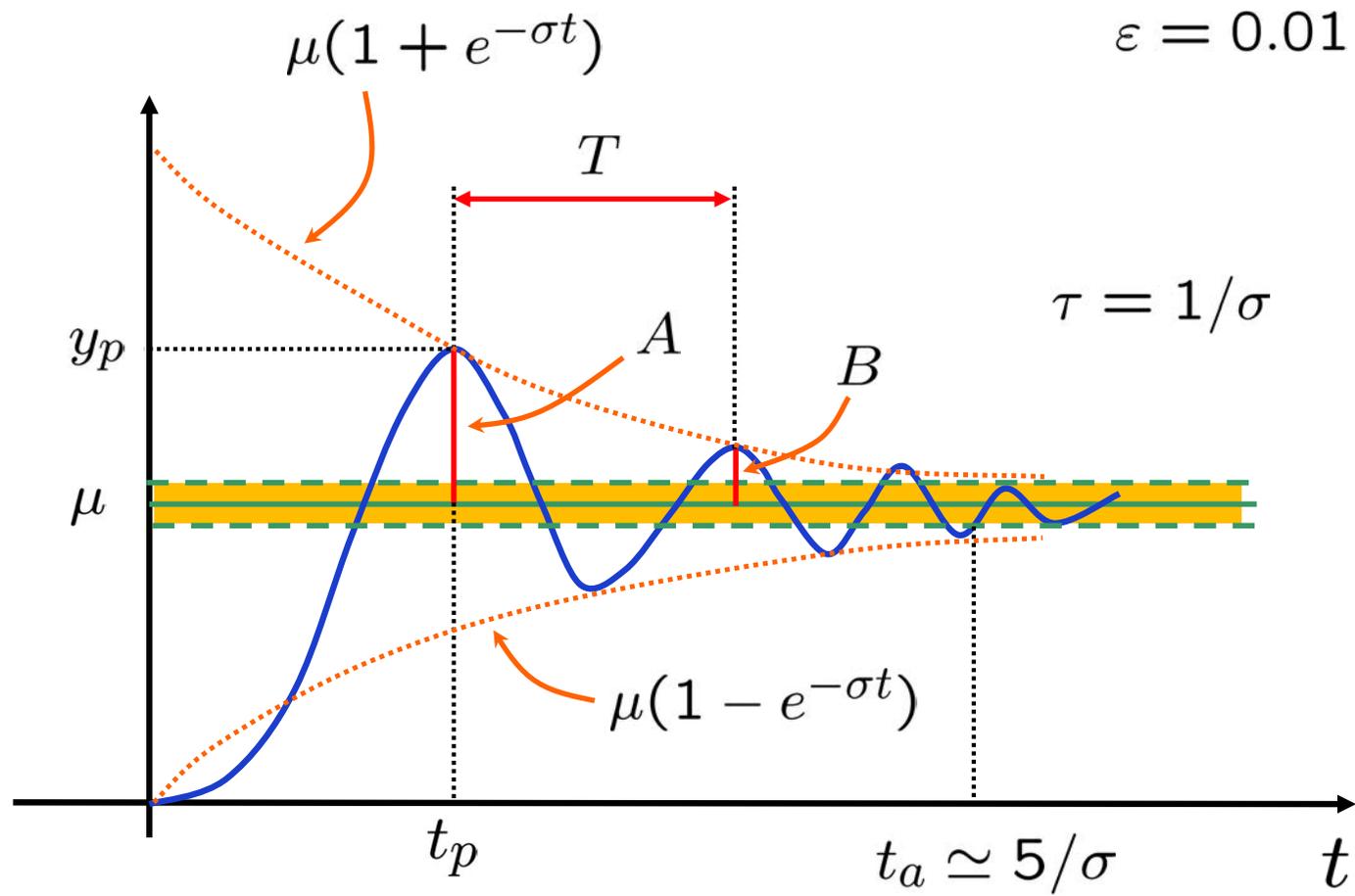
$$\dot{y}(0) = 0$$

$$\ddot{y}(0) = \rho > 0$$

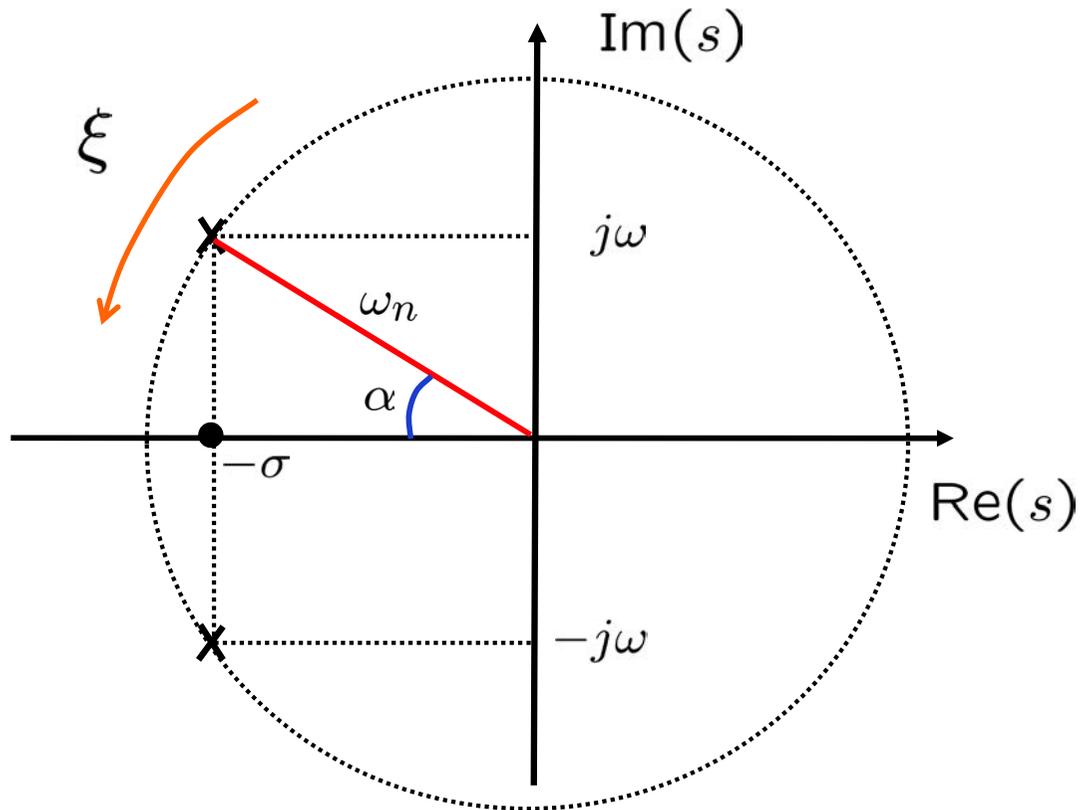
Per esercizio dimostrare che si può riscrivere la risposta così:

$$y(t) = \mu \left[1 - \frac{\sqrt{\sigma^2 + \omega^2}}{\omega} e^{-\sigma t} \sin(\omega t + \varphi) \right], \quad t \geq 0$$

Dove: $\varphi = \arccos \left(\frac{\sigma}{\sqrt{\sigma^2 + \omega^2}} \right)$



- Diversa parametrizzazione



$$\omega_n^2 = \sigma^2 + \omega^2$$

$$\omega_n \xi = \sigma$$

$$\omega_n \sqrt{1 - \xi^2} = \omega$$

ω_n

Pulsazione naturale

$$\cos \alpha = \xi$$

Smorzamento

$$0 \leq \xi \leq 1$$



$$G(s) = \frac{q}{s^2 + 2\xi\omega_n s + \omega_n^2}$$

- Espressione della FdT

$$G(s) = \frac{\rho}{(s + \sigma + j\omega)(s + \sigma - j\omega)} = \frac{\rho}{(s + \sigma)^2 + \omega^2}$$

$$= \frac{\rho}{s^2 + \underbrace{2\sigma s}_{2\xi\omega_n} + \underbrace{\sigma^2 + \omega^2}_{\omega_n^2}} = \frac{\rho}{s^2 + 2\xi\omega_n s + \omega_n^2}$$

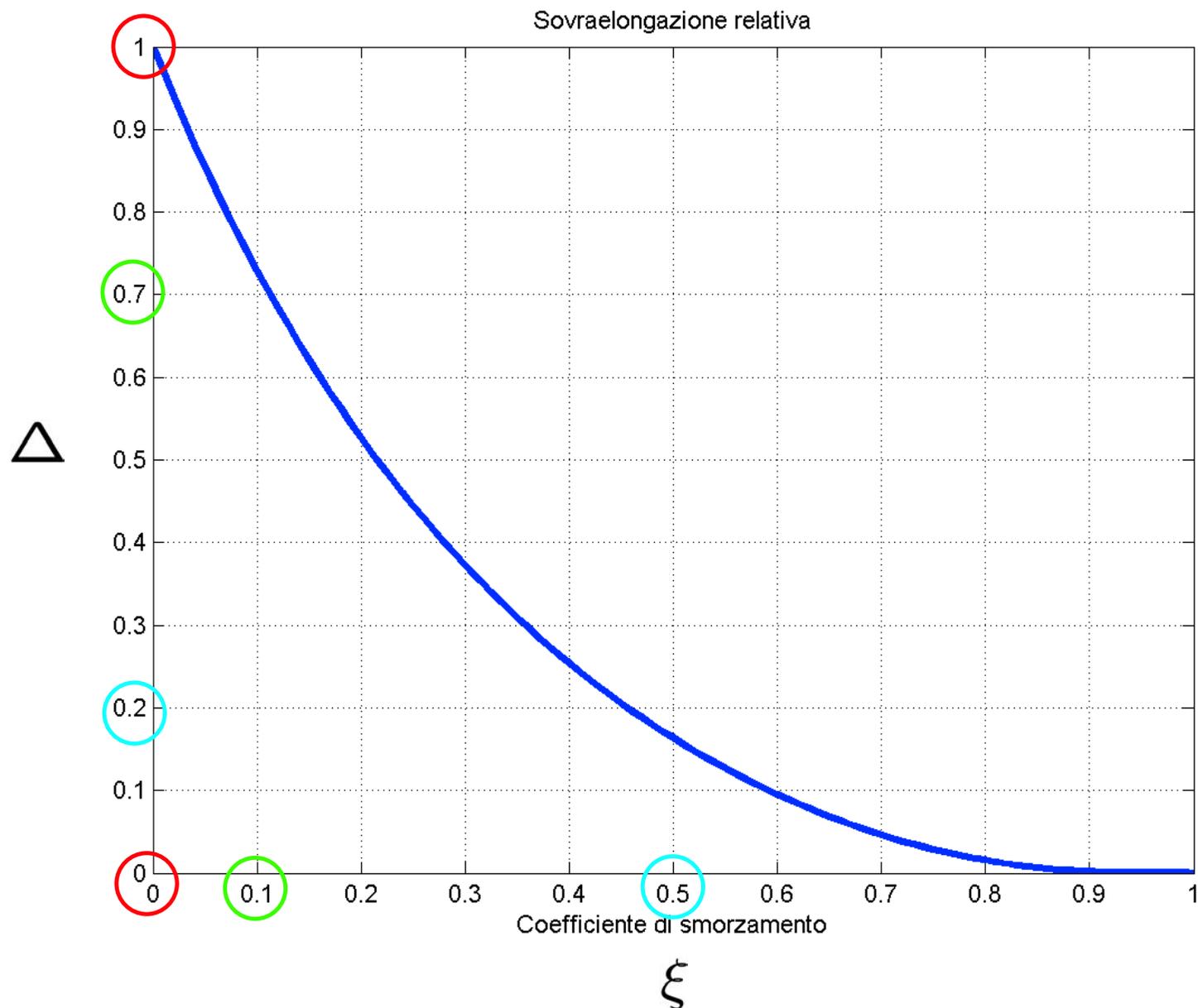
$$= \frac{\rho/\omega_n^2}{1 + \frac{2\xi}{\omega_n}s + \frac{1}{\omega_n^2}s^2} = \frac{\mu}{1 + \frac{2\xi}{\omega_n}s + \frac{1}{\omega_n^2}s^2}$$

dove $\mu := \frac{\rho}{\omega_n^2}$

- Parametri caratteristici

- $t_a \simeq \frac{5}{\sigma} = \frac{5}{\xi\omega_n}$
- $T = \frac{2\pi}{\omega} = \frac{2\pi}{\omega_n\sqrt{1-\xi^2}}$
- $t_p = \frac{\pi}{\omega} = \frac{\pi}{\omega_n\sqrt{1-\xi^2}}$
- $y_p = \mu \left[1 + e^{-\frac{\sigma\pi}{\omega}} \right] = \mu \left[1 + e^{-\frac{\xi\pi}{\sqrt{1-\xi^2}}} \right]$
- $\Delta = \frac{A}{\mu} = e^{-\sigma\pi/\omega} = e^{-\frac{\xi\pi}{\sqrt{1-\xi^2}}}$
- $\frac{B}{A} = \dots = \Delta^2 = e^{-2\sigma\pi/\omega} = e^{-\frac{2\xi\pi}{\sqrt{1-\xi^2}}}$

Dipendono solo da ξ
e non da ω_n



- Casi limite

- $\xi = 0$ $G(s) = \frac{\rho}{s^2 + \omega_n^2}$ Poli: $\pm j\omega_n$



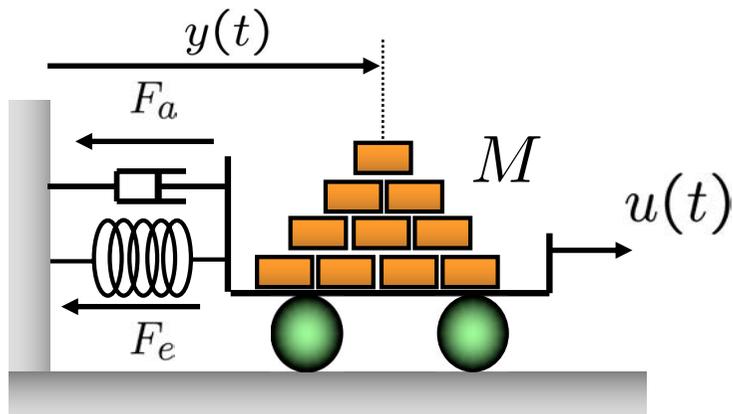
Oscillazioni non smorzate

- $\xi = 1$ $G(s) = \frac{\rho}{(s + \omega_n)^2}$ Poli: $-\omega_n, -\omega_n$



Assenza di oscillazioni

- Esempio 1



$$G(s) = \frac{1}{Ms^2 + hs + k}$$

$$= \frac{1/M}{s^2 + \frac{h}{M}s + \frac{k}{M}}$$

$$s^2 + 2\xi\omega_n s + \omega_n^2$$

$$\mu = G(0) = \frac{1}{k}$$

$$2\xi\omega_n = \frac{h}{M} \quad \omega_n^2 = \frac{k}{M}$$

$$T = \frac{2\pi}{\omega_n\sqrt{1-\xi^2}}$$

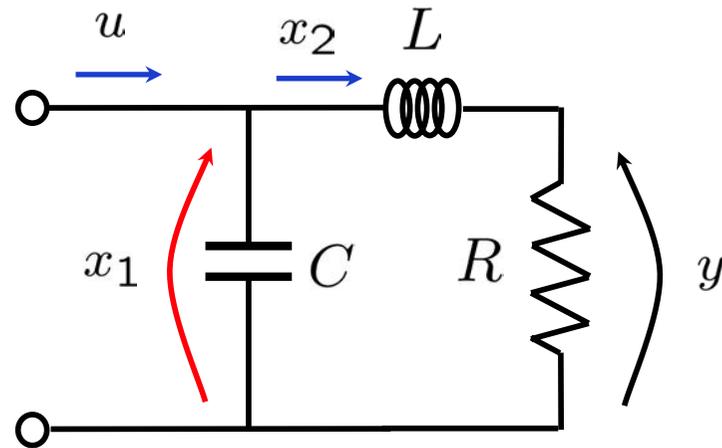
$$\omega_n = \sqrt{\frac{k}{M}}$$

$$\Delta = e^{-\frac{\xi\pi}{\sqrt{1-\xi^2}}}$$

$$\xi = \frac{h}{2\sqrt{kM}}$$



- Esempio 2



$$\begin{cases} C\dot{x}_1 = u - x_2 \\ L\dot{x}_2 = x_1 - Rx_2 \\ y = Rx_2 \end{cases}$$

$$\downarrow \quad A = \begin{bmatrix} 0 & -1/C \\ 1/L & -R/L \end{bmatrix} \quad B = \begin{bmatrix} 1/C \\ 0 \end{bmatrix} \quad C = [0 \quad R]$$

$$\downarrow \quad G(s) = [0 \quad R] \begin{bmatrix} s & 1/C \\ -1/L & s + R/L \end{bmatrix}^{-1} \begin{bmatrix} 1/C \\ 0 \end{bmatrix} = \dots = \frac{R/(LC)}{s^2 + \frac{R}{L}s + \frac{1}{LC}}$$

$$\downarrow \quad \omega_n = \frac{1}{\sqrt{LC}} \quad \xi = \frac{R}{2} \sqrt{\frac{C}{L}} \quad \mu = R$$

D)

$$G(s) = \frac{\mu(1 + sT)}{1 + \frac{2\xi}{\omega_n}s + \frac{1}{\omega_n^2}s^2}$$

$$0 < \xi < 1$$

$$\omega_n > 0$$

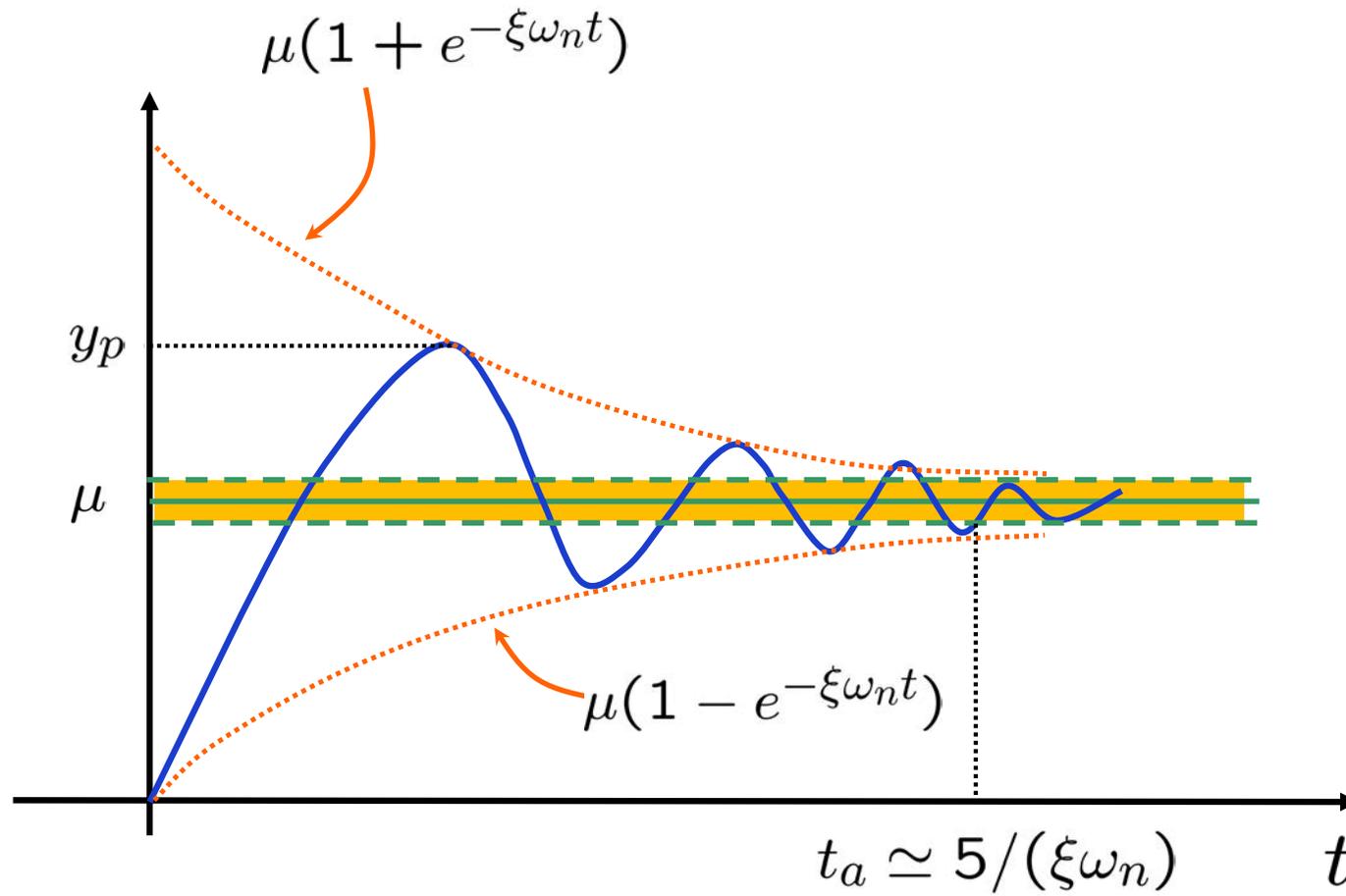
$$\mu > 0$$

Alcune caratteristiche della risposta:

$$y(\infty) = \mu > 0$$

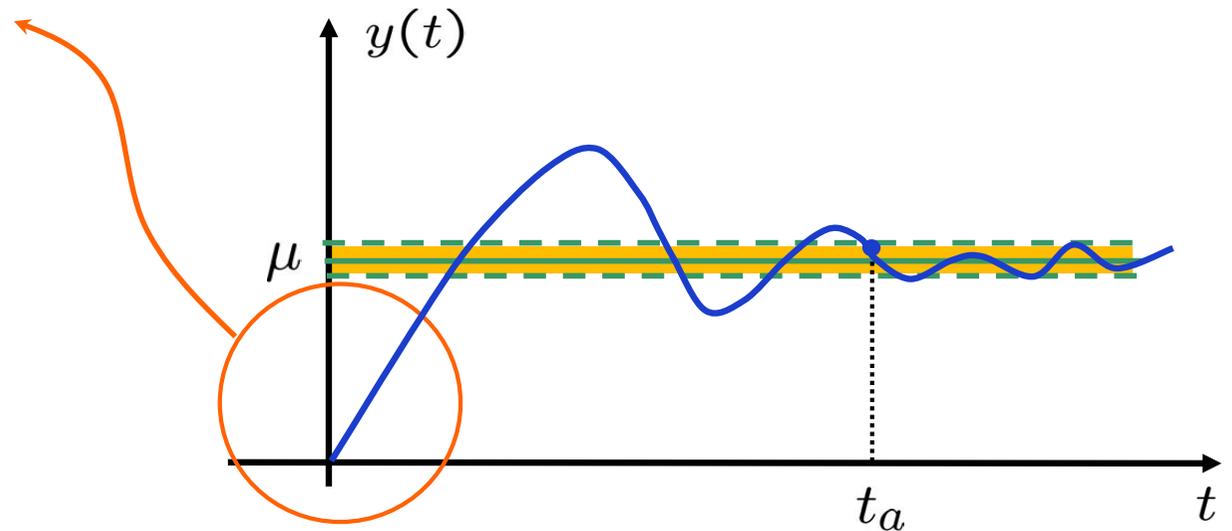
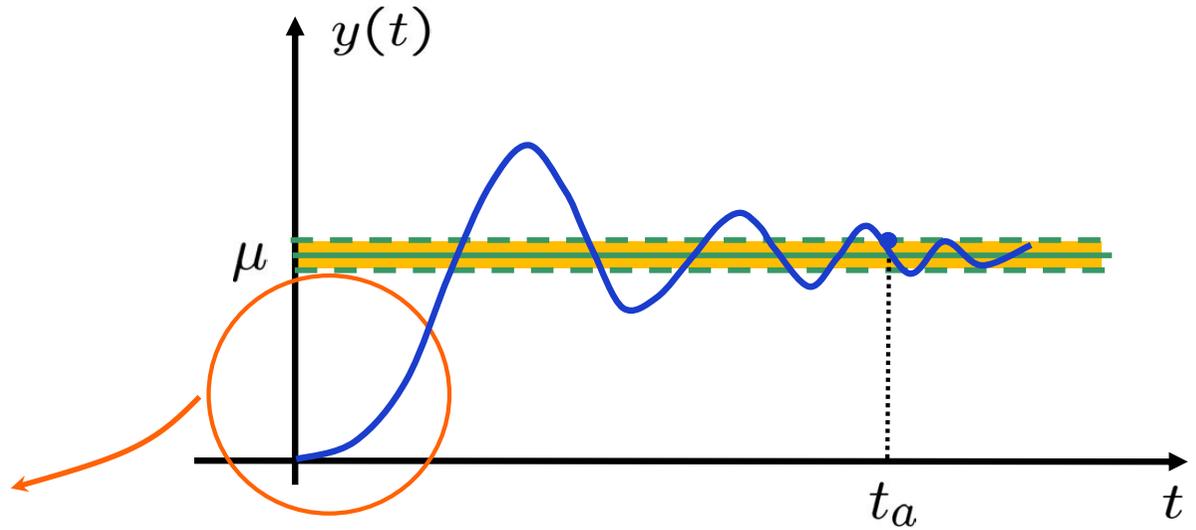
$$y(0) = 0$$

$$\dot{y}(0) = \mu T \omega_n^2 \begin{cases} > 0, & \text{se } T > 0 \\ < 0, & \text{se } T < 0 \end{cases}$$



- Confrontando C) e D):

A parità di tempo di assestamento e' diverso il tempo di salita, ovvero la velocita' di risposta



- Risposta allo scalino – Ordine > 2

$$G(s) = \frac{\mu}{s^g} \frac{\prod_{i=1}^m (1 + sT_i)}{\prod_{i=1}^n (1 + s\tau_i)}$$

As. stabilita`



Re (poli) < 0

$g \leq 0$

salvo cancellazioni

$$y(t) = \mathcal{L}^{-1} \left[G(s) \frac{1}{s} \right]$$

- Teo. valore iniziale

$$\lim_{t \rightarrow 0^+} y(t) = \lim_{s \rightarrow \infty} \cancel{s} \frac{1}{\cancel{s}} G(s) \quad \begin{array}{l} = 0, \quad \text{se } m < n \\ \neq 0, \quad \text{se } m = n \end{array}$$

- Teo. valore finale

$$\lim_{t \rightarrow \infty} y(t) = \lim_{s \rightarrow 0} \cancel{s} \frac{1}{\cancel{s}} G(s) \quad \begin{array}{l} = \mu, \quad \text{se } g = 0 \\ = 0, \quad \text{se } g < 0 \end{array}$$

- Approssimazione a poli dominanti

- Caso semplice – Poli reali distinti

$$Y(s) = G(s) \frac{1}{s} = \frac{\alpha_0}{s} + \frac{\alpha_1}{1 + s\tau_1} + \dots + \frac{\alpha_n}{1 + s\tau_n}$$

↳

$$y(t) = \mathcal{L}^{-1} \left[G(s) \frac{1}{s} \right]$$

$$= \alpha_0 + \frac{\alpha_1}{\tau_1} e^{-t/\tau_1} + \dots + \frac{\alpha_n}{\tau_n} e^{-t/\tau_n}$$

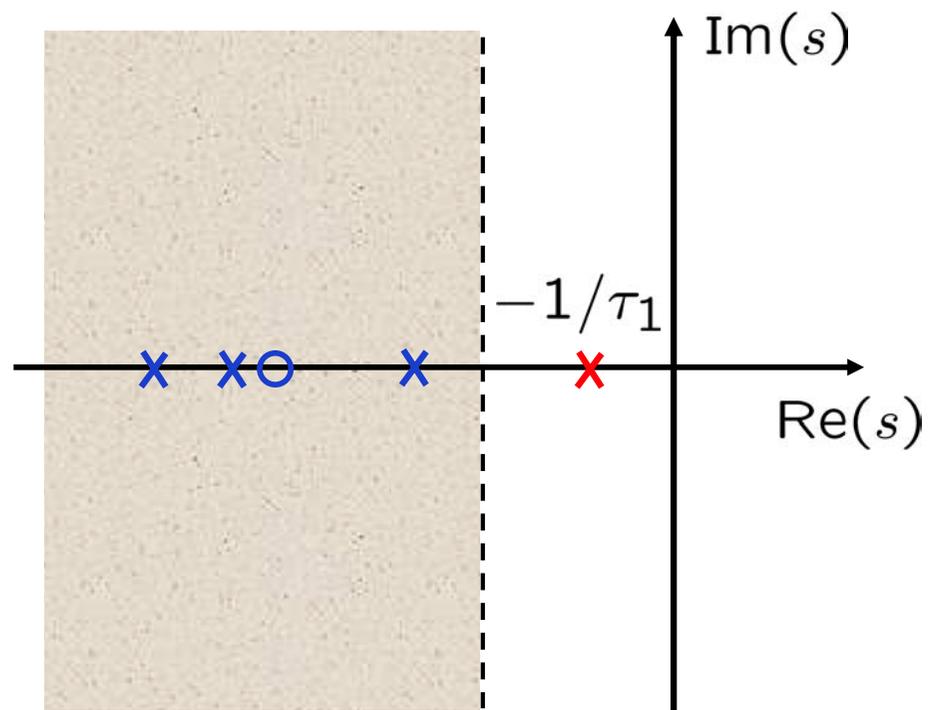
Se $\tau_1 > \tau_2 > \dots > \tau_n$

↳

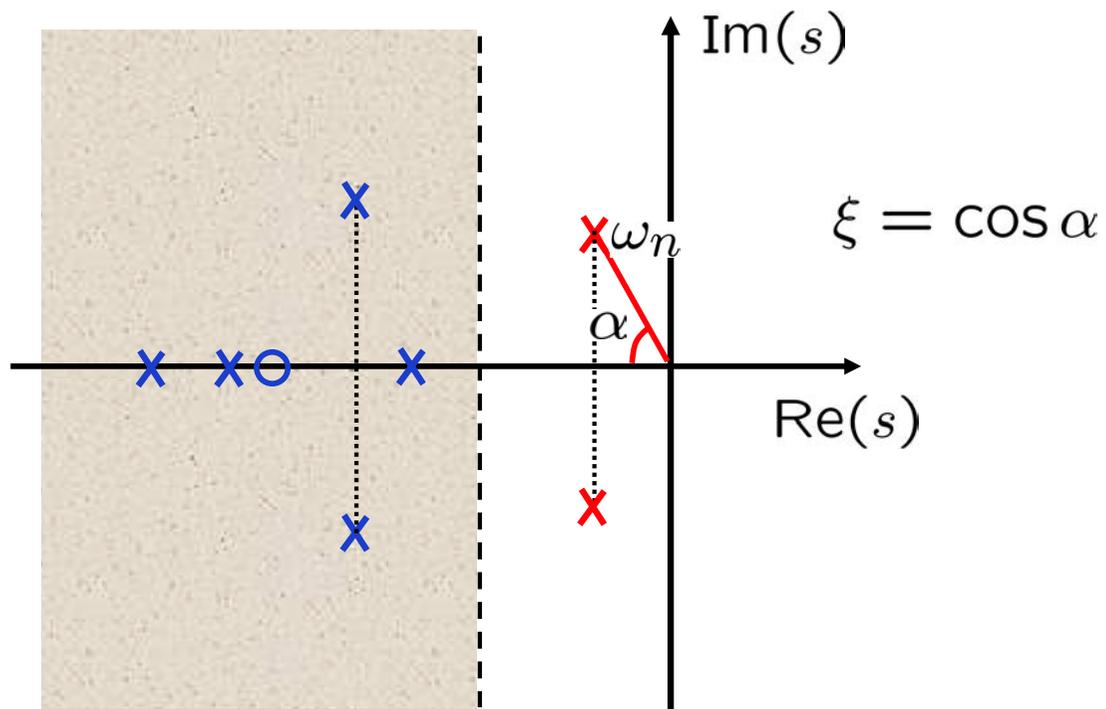
$$y(t) = \alpha_0 + \frac{\alpha_1}{\tau_1} e^{-t/\tau_1} + \dots + \frac{\alpha_n}{\tau_n} e^{-t/\tau_n}$$

$$\simeq \alpha_0 + \frac{\alpha_1}{\tau_1} e^{-t/\tau_1} \quad \text{componente dominante} \quad t_a \simeq 5\tau_1$$

- Polo dominante reale



- Poli dominanti complessi



- AVVERTENZA



Nell'approssimazione a poli dominanti e' necessario

- Tener conto di zeri "vicini" all'asse immaginario
- Conservare il guadagno μ

- Esempio

$$G(s) = \frac{400(1 + s)}{(1 + 0.2s)(1 + 0.1s)(s^2 + 2s + 4)}$$

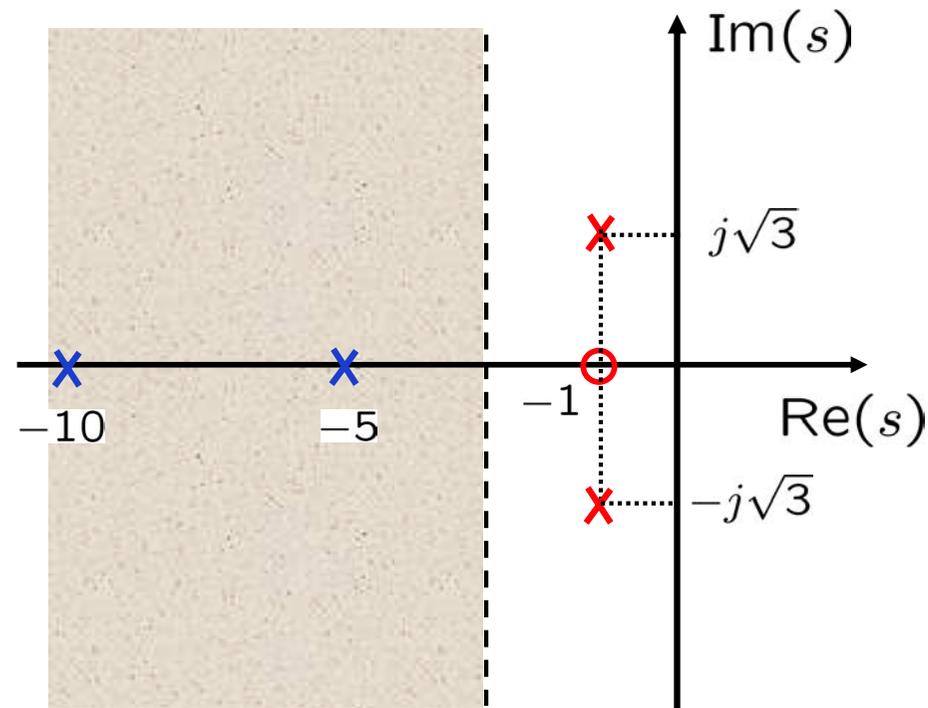
$$\mu = G(0) = 100$$

Poli: -5

-10

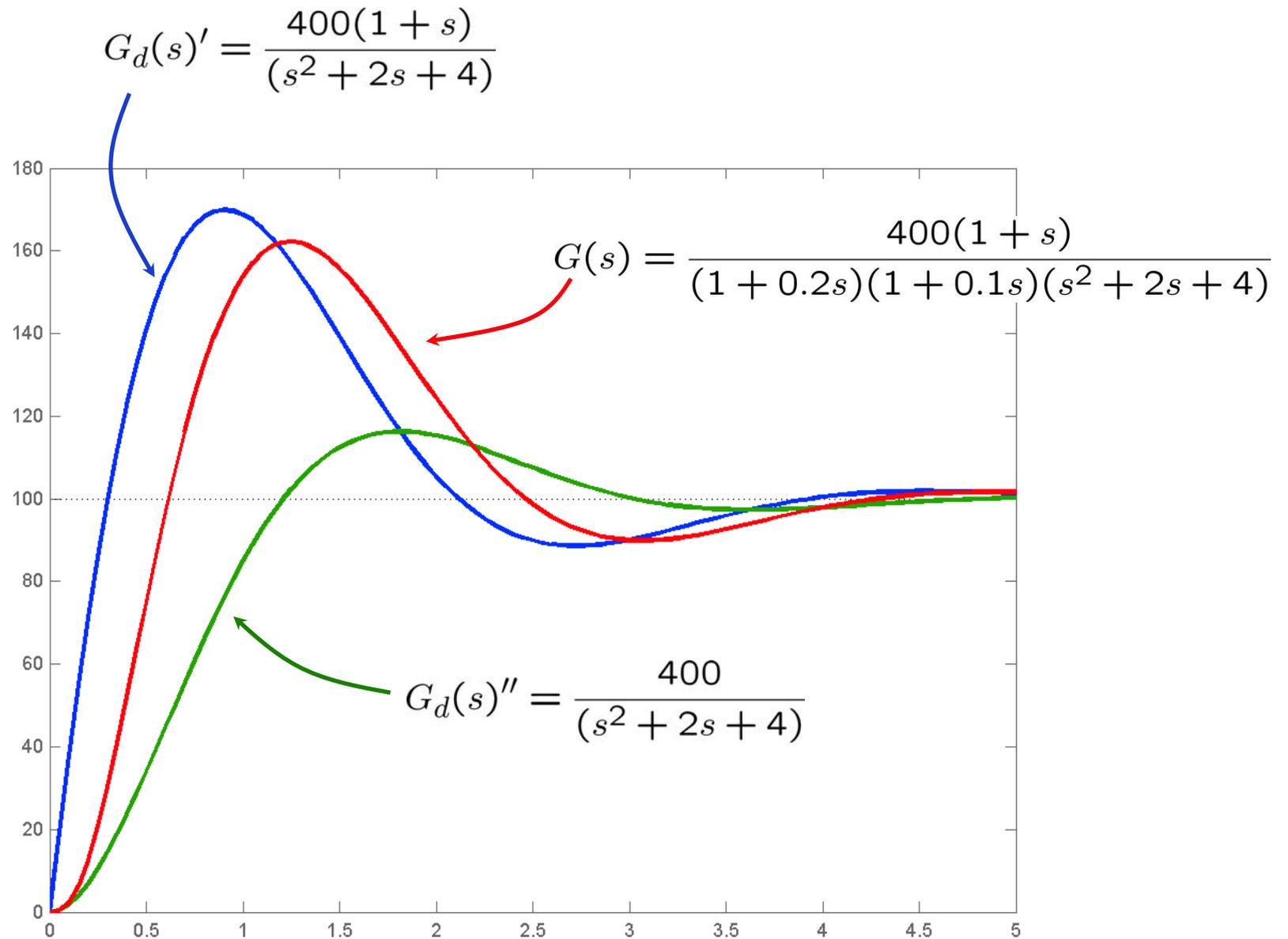
$-1 \pm j\sqrt{3}$

Zero: -1



$$\omega_n = 2$$

$$\xi = 1/2$$



- Approssimazione a polo equivalente

–Hp: poli reali negativi

$$G(s) = \frac{\mu}{(1 + s\tau_1)(1 + s\tau_2) \cdots (1 + s\tau_n)} \quad \tau_i > 0$$

$$G_e(s) = \frac{\mu}{1 + s\tau_e}, \quad \tau_e := \sum_{i=1}^n \tau_i$$

Componente dominante $t_a \simeq 5\tau_e$

- Esempio

$$G(s) = \frac{1}{(1+s)(1+0.1s)(1+0.5s)}$$

$$\mu = G(0) = 1$$

Poli: -1
-2
-10

