

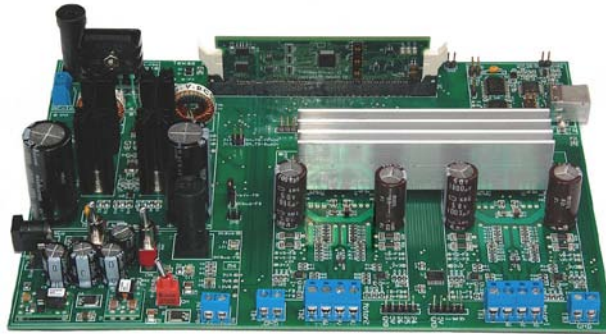
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# F2833x Digital Motor Control

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## Introduction

In this module, we will look into an application that is not usually considered to be the domain of Digital Signal Processors: real-time control of electrical motors. In the old days, the control of speed and torque of electrical motors was performed using purely analog technology. Since the appearance of microprocessors, more and more control units have been designed digitally, using the advantages of digital systems. This improves the degree of efficiency and allows the implementation of more advanced control schemes, thanks to increased real-time computing power. It is a natural progression to use the internal hardware computing units of a DSP to transfer the calculation from a standard microprocessor to a DSP. This way, we can implement more advanced algorithms in a given time period.



However, to use a digital controller for motor control, the system needs a little more than computing power. The output signals of the digital controller to the power electronic are usually generated as pulse width modulated signals (PWM). It would be most cost-effective if the controller could be equipped with an internal PWM-unit. To control the operation of the motor we need to do some measurements for currents and voltages – analogue to digital converters (ADC) will be helpful as well. A typical unit to perform a position/speed measurement is an optical encoder; quite often, we build in a Quadrature Encoder (QEP). Recalling all parts of the F2833x we discussed in this Teaching - CD, you can imagine that the F2833 is an ideal device for Digital Motor Control (DMC).

The chapter will not go into the fine details of electrical motors and drives. Instead, it will give you a sense of what needs to be done to use the F2833x to control the motor of a vacuum cleaner or the motor of an electrical vehicle. To fully understand these principles, it requires many more classes at university. If you are on a course of electrical engineering that focuses on drives and power engineering, you might be familiar with most of the technical terms. If not, see this chapter as a challenge for you to open up another application field for a Digital Signal Controller.

Chapter 18 is based on a Texas Instruments Presentation “TIs C2000 Real-Time MCU for Digital Motor Control“(August 2009). Depending on the laboratory equipment at your university, you might be offered the chance to attend a laboratory session to build a working solution for such a motor control.

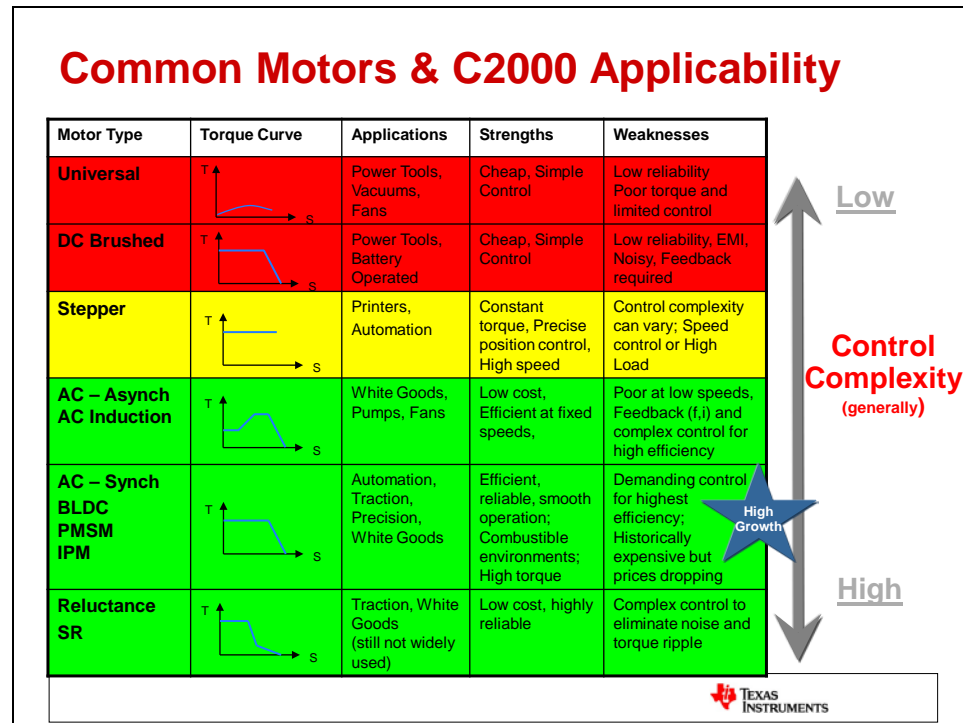
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# Basics of Electrical Motors

## Motor Categories

In order to classify the different electrical motors families, we can distinguish motors driven by direct current (DC) and motors driven by an alternating current (AC). DC motors are the most popular ones: both stators and rotors carry an excitation created by coils or windings in which DC current circulates. In order to ensure the motor rotation by commutating the windings, brushes are permanently in contact with the rotor.

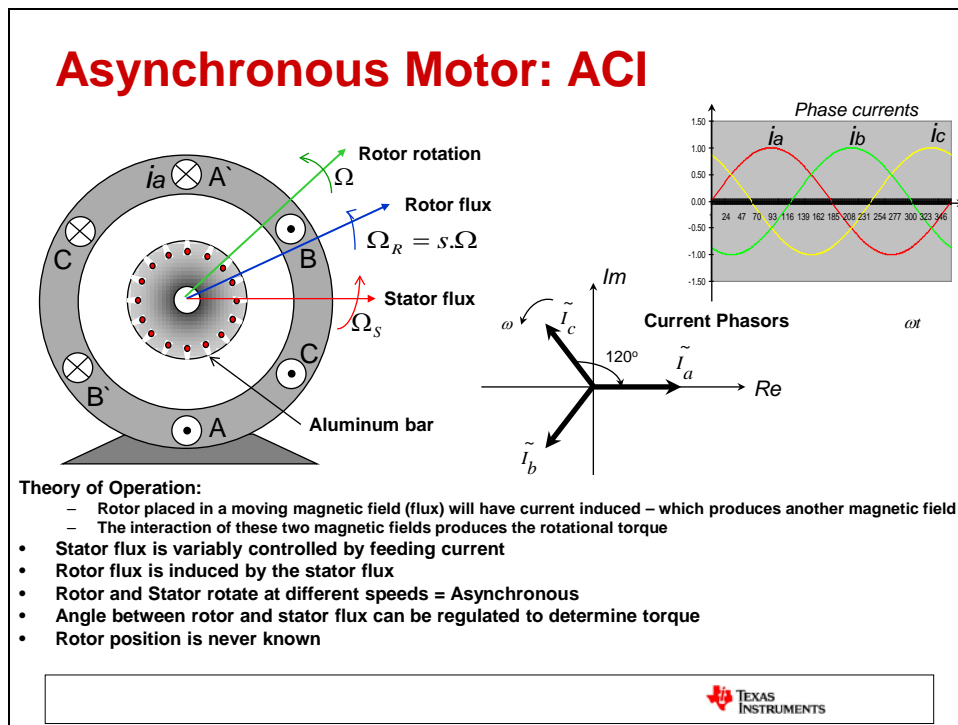


Under the classification of AC motors, we have synchronous motors and asynchronous motors; both motor types are induction machines.

Asynchronous machines require a sinusoidal voltage distribution on the stator phases in order to induce current on the rotor, which is not fed by any currents nor carries any magnetic excitation.

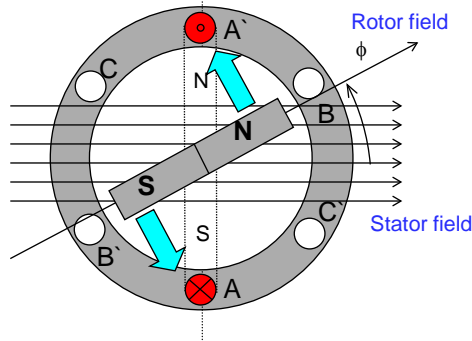
Synchronous motors are usually called “Brushless DC Motors” (BLDC) but can also be called “Permanent Magnet Synchronous Motors” (PMSM) depending on their construction and the way they are being controlled. In this type of motor, we have one sinusoidal or trapezoidal excitation on the stator and one constant flux source on the rotor (usually a magnet).

## Asynchronous Motor



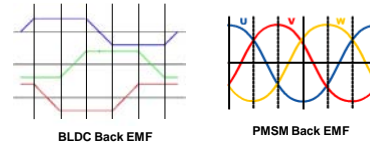
# Synchronous Motors: BLDC and PMSM

## Synchronous Motors: BLDC & PMSM



$$\text{Rotor speed (rad/s): } \Omega = \frac{\omega}{p} \text{ gives } \frac{120 \cdot f}{p} \text{ (r.p.m.)}$$

$f$  : AC supply frequency (Hz)  
 $p$  : motor poles



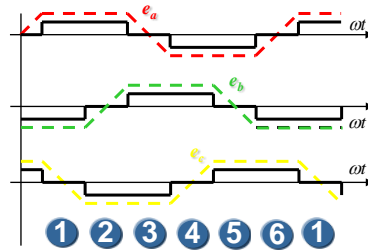
### Theory of Operation:

- Fixed rotor flux (magnetic field) and a produced stator flux
- The interaction between the two fields produces a torque which will cause the motor to rotate
- Stator flux is variably controlled by feeding current
- Rotor flux is constant by permanent magnets or current fed coils
- Rotor rotation is at same frequency as supplied excitation = Synchronous
- Angle between rotor and stator flux can be regulated to determine torque
- Rotor position can be measured or estimated



## Which Synchronous? BLDC vs. PMSM

Back EMF of BLDC Motor

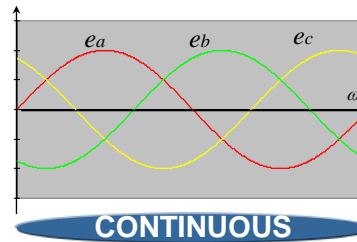


### • BLDC Motors

- Easier to control (6 Trapezoidal states)
- Torque ripple at commutations
- Better for lower speed
- Noisy
- Doesn't work with distributed winding
- Not as efficient, lower Torque
- Lower cost

### Continuous States = More Complex

Back EMF of PMSM



### • PMSM Motors

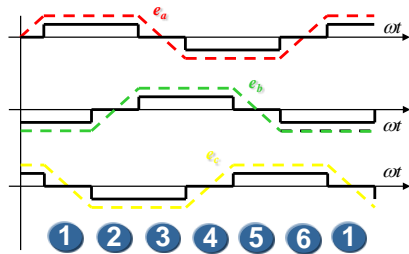
- More complex control (continuous 3Ph Sine Wave)
- No torque ripple at commutation
- Higher max achievable speed
- Low noise
- Work with low-cost distributed winding
- Higher efficiency, higher Torque
- Higher cost



# Motor Control Principles

## Trapezoidal Control

### Trapezoidal Control (BLDC only) - Simple




**Trapezoidal Control**

- + Fed with direct current
- + Stator Flux commutation only each 60° (1-6)
- + Two phases ON at the same time
- Torque ripple at commutations
- Commutation at high speed difficult
- Noisy
- Doesn't work with distributed winding

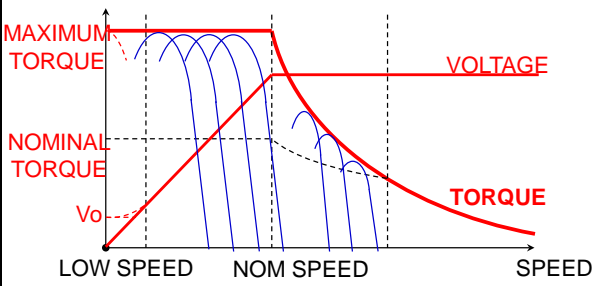
**Trapezoidal Step by Step**

1. Sample the System
  - Position: Hall sensors (3 Square wave outputs correspond to rotor position)
  - 3 Ph Voltages (sensorless)
  - Current through a shunt (optional)
2. Calculate
  - Speed = Frequency of positions
3. Estimations (Sensorless)
  - Voltage Back EMF easily gives position of rotor
4. Regulate the Loop
  - Simple proportional PID speed control
  - Optional Current/Voltage Monitor
5. Stimulate the System
  - Simple PWM trapezoidal patterns (1-6)

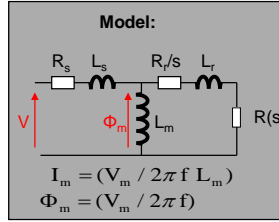


## Scalar Control (“V/f”)

### V/f Control (ACI or PMSM) – Simple



**Model:**

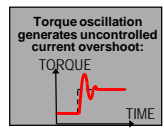



$$I_m = (V_m / 2\pi f L_m)$$

$$\Phi_m = (V_m / 2\pi f)$$

**At low speed:**  $R_s$  is no longer negligible:  $V_m < V$   
A large portion of energy is now wasted.

- + Simple Proportional Control: Three sine waves feeding the motor
- + Position information not required (optional)
- Poor dynamic performance
- Torque delivery not optimized for all speeds





The V/Hz regulation scheme is the simplest one that can be applied to an **asynchronous motor**. The goal is to work in an area where the rotor flux is constant (Volts proportional to speed).

In practical solutions, the speed sensor is optional as the control is tuned to follow a predefined “speed-profile versus load table”, assuming the load characteristics are known in advance.

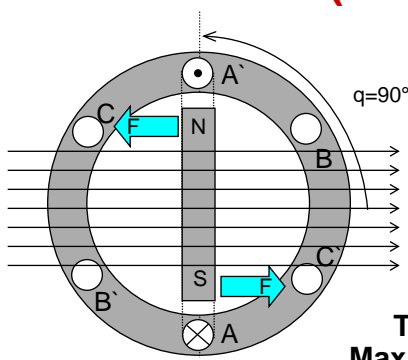
Obviously, this type of control bases itself on the steady electrical characteristics of the machine and assumes that we are able to work with a constant flux in the complete speed range the application targets. This is why this type of control does not deliver a good dynamic performance and a good transient response time; the V/Hz profile is fixed and does not take into account conditions other than those seen in a steady state. The second point is the problem at startup of AC induction motors, which cannot deliver high torques at zero speed; in this case, the system cannot maintain a fixed position. In practice for low speed, we need to increase the delivered voltage to the stator compared to the theoretical V/Hz law.

## Field Oriented Control (FOC)

Instead of using a pure sine wave shaped modulation of the PWM stage, in recent years the space vector theory has demonstrated some improvements for both the output crest voltage and the harmonic copper loss. The maximum output voltage based on the space vector theory is 1.155 times larger than the conventional sinusoidal modulation. This makes it possible to feed the motor with a higher voltage than the simpler sub-oscillation modulation method. This modulator enables higher torque at high speeds, and a higher efficiency. Torque distortion is also reduced.

The space vector PWM technique implemented into the existing TI DMC library reduces the number of transistor commutations. It therefore improves EMI behavior.

## FOC Control (ACI or PMSM) – Complex




Maintain  
the 'load  
angle' at  
90°

**Torque = Cross Product of Fields  
Maximized when  $\sin q = 1 = 90^\circ$  Degrees**

**Field Orientation**

- + Reduced torque ripple
- + Better dynamic response
- + Good performance at lower speeds
- Need to measure angle between rotor and stator
- Requires independent control of flux and torque in real-time

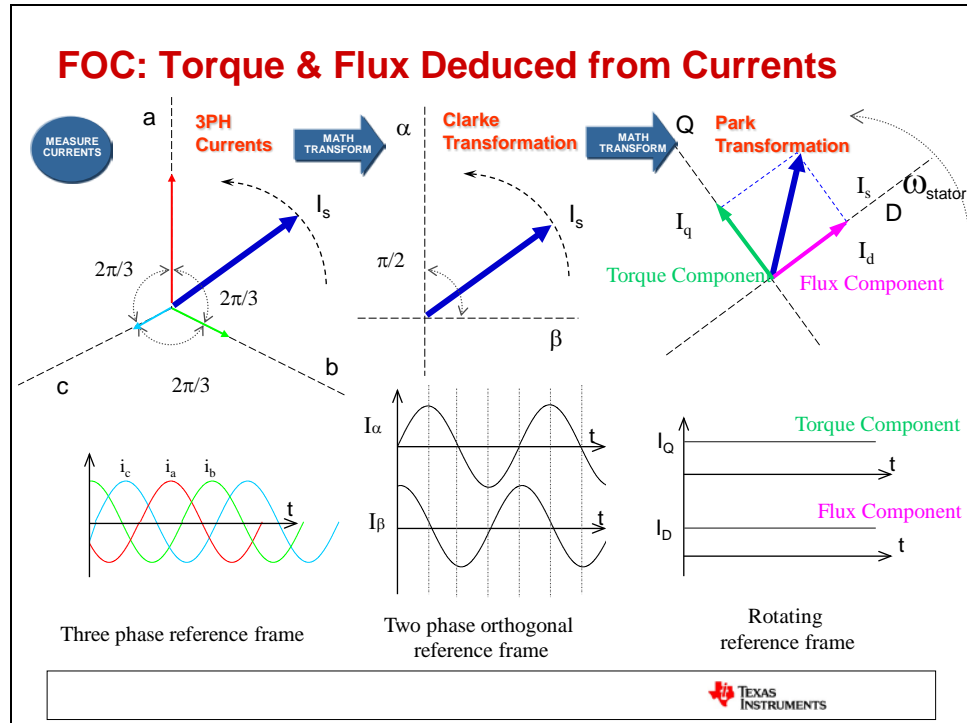
**Measure & Control**  
 Torque, Flux  
 Angle, Speed



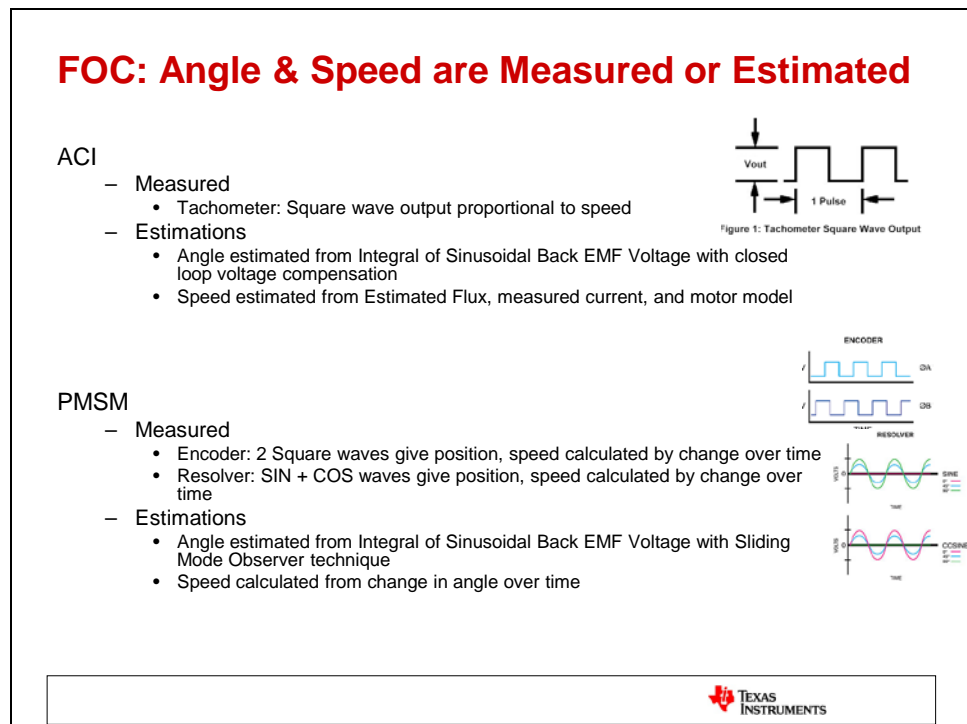
A typical characteristic of FOC - PWM command strategy is that the envelope of the generated signal is carrying the first and the third harmonics. We can interpret this as a consequence of the special PWM sequence applied to the power inverters. Literature also mentions the third harmonic injection to boost out the performance we get out of the DC bus capacitor. This third-harmonic exists in the phase to neutral voltage but disappears in the phase-to-phase voltage.



## FOC Coordinate Transform (Clarke / Park)



## FOC Measurement of Motor Position and Speed



## Advantages of Vector Control

### Why FOC? Performance Comparison

Performance Comparison Control Algorithms	Volts per Hertz Control	FOC	
		Vector Drive	Servo Drive
Velocity Loop bandwidth in Hz	1Hz	50 Hz	100 Hz
Minimum speed with full load (RPM)	90	0	0
Maximum speed with 25% load (RPM)	1.5 X Base Speed	2.5 X Base Speed	2.0 X Base Speed
Minimum acceleration time (seconds)	3	0.1	0.01
Minimum deceleration time (seconds)	3	0.1 w/DB	0.01
Maximum starting torque (%)	150%	200%	200%
Speed regulation at full load (%)	± 3% (base)	± 0.01% (set)	± 0.01% (set)

The numbers above are for comparison only. Specific system dynamics will affect exact data.

#### FOC

- System responds faster to changes in set point or load change
- Minimum speed at full load is now essentially zero
- Starting torque is increased
- Very little torque ripple
- Reduces Cost
  - Optimally size motor for the task at hand
  - Current controlled, so the inverter can be optimized



## FOC Step By Step

### FOC: Step By Step

1. Sample the System
  - Current
  - Voltage (Sensorless)
  - Speed (ACI Sensored)
  - Speed & Position/Angle (PMSM Sensored)
2. Transform sampled data and calculate useful quantities
  - Measure Currents → Use Clarke/Parke Transform → Torque & Flux
  - ACI Sensored
    - Measure Speed + Torque & Flux Components → Angle
  - ACI Sensorless
    - Measure Voltage + Stator Current (Clarke) → Angle Estimation
    - Angle Estimation + Stator Current (Clarke) → Speed Estimation
  - PMSM Sensored
    - Measured Speed and Measured Angle
  - PMSM Sensorless
    - Measured Voltage + Stator Current (Clarke) → Angle Estimation
    - Angle Estimation → Speed Estimation
3. Regulate the Loop
  - PID techniques are most common; Controls to a reference value
  - Regulate speed, position/angle, current, flux and maximize torque
4. Stimulate the System
  - Inverse transforms of Park and Clarke
  - PWM Pattern generation to drive the voltage/current source

Measure & Control  
Torque, Flux  
Angle, Speed

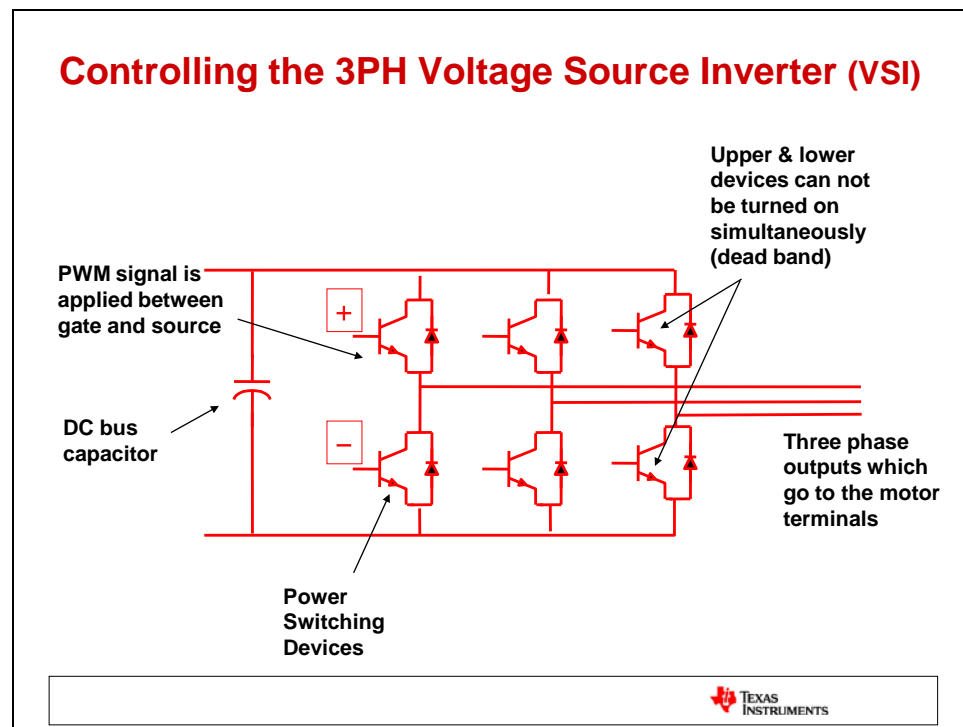


## 3-Phase Power Switches

As we saw in the previous basic diagrams, we need to apply three 120° phase shifted excitation signals to the power circuitry of the motor. As you have seen in Chapter 7 (“PWM”), a PWM signal can be used to modulate sinusoidal signals. With three independent switching pattern streams and six power switches, we can deliver the necessary phase voltages to generate the required torque imposed by the load. The goal is to build the correct conduction sequences in the IGBTs to deliver sinusoidal currents to the motor to transform it to a mechanical rotation.

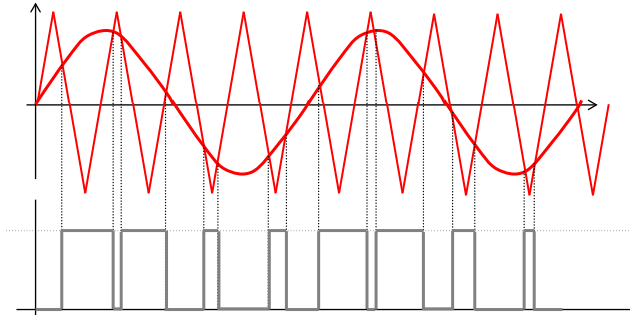
This is traditionally achieved by comparing a three-phase sinusoidal waveform with a triangular carrier. In the digital world, on the DSP processor, we compute a sinusoidal command and apply it to the PWM units that generate the appropriate PWM outputs usually connected to gate drivers of the IGBTs from the inverter.

Basically we are “chopping” a DC voltage, carried by the DC bus capacitor, in order to build the appropriate voltage shapes for the stator phases, with the goal of having a good efficiency during this energy conversion process. This is a power electronics concern: we need to minimize the noise introduced by these conducting sequences and source of harmonics.



## Sine PWM VSI Control

### Traditional (Old) Sine PWM VSI Control

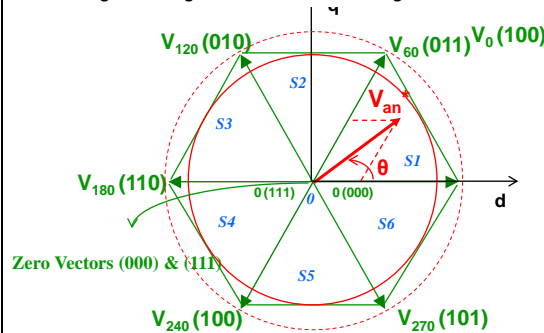


- ◆ **Inputs**
  - ◆ Triangular Switching Frequency (5-25 kHz typically)
  - ◆ Sine wave = Carrier Trying to match (V or I Reference, 0-1000 Hz typically)
  - ◆ Image not to scale; Typically 100s of triangle periods in each Sine wave
- ◆ **Output**
  - ◆ When they cross, you switch the PWM

## Space Vector PWM VSI Control

### Space Vector PWM VSI Control

Theory: A special switching sequence of the upper three power devices of a VSI results in 3 pseudo-sinusoidal currents in the stator phases. Calculate the appropriate duty cycle – every period - needed to generate a given stator reference voltage.



For review of the calculations, please see detailed theory in SVGEN documentation from DMC Library (SPRC080 or SPRC125).

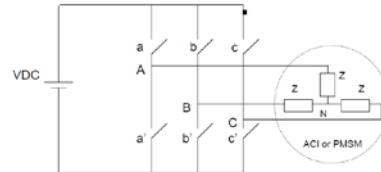


Figure 2: Power bridge for a three-phase VSI

c	b	a	$V_{an}$	$V_{bn}$	Vector
0	0	0	0	0	$O_n$
0	0	1	$\frac{2}{3}V_{DC}$	0	$U_0$
0	1	0	$-\frac{V_{DC}}{3}$	$\frac{V_{DC}}{\sqrt{3}}$	$U_{120}$
0	1	1	$\frac{V_{DC}}{3}$	$\frac{V_{DC}}{\sqrt{3}}$	$U_{180}$
1	0	0	$-\frac{V_{DC}}{3}$	$-\frac{V_{DC}}{\sqrt{3}}$	$U_{240}$
1	0	1	$\frac{V_{DC}}{3}$	$-\frac{V_{DC}}{\sqrt{3}}$	$U_{270}$
1	1	0	$-\frac{2}{3}V_{DC}$	0	$U_{180}$
1	1	1	0	0	$O_{111}$

Table 2: Switching patterns, corresponding space vectors and their (a,b) components



### SVPWM Benefits vs Traditional Sine PWM

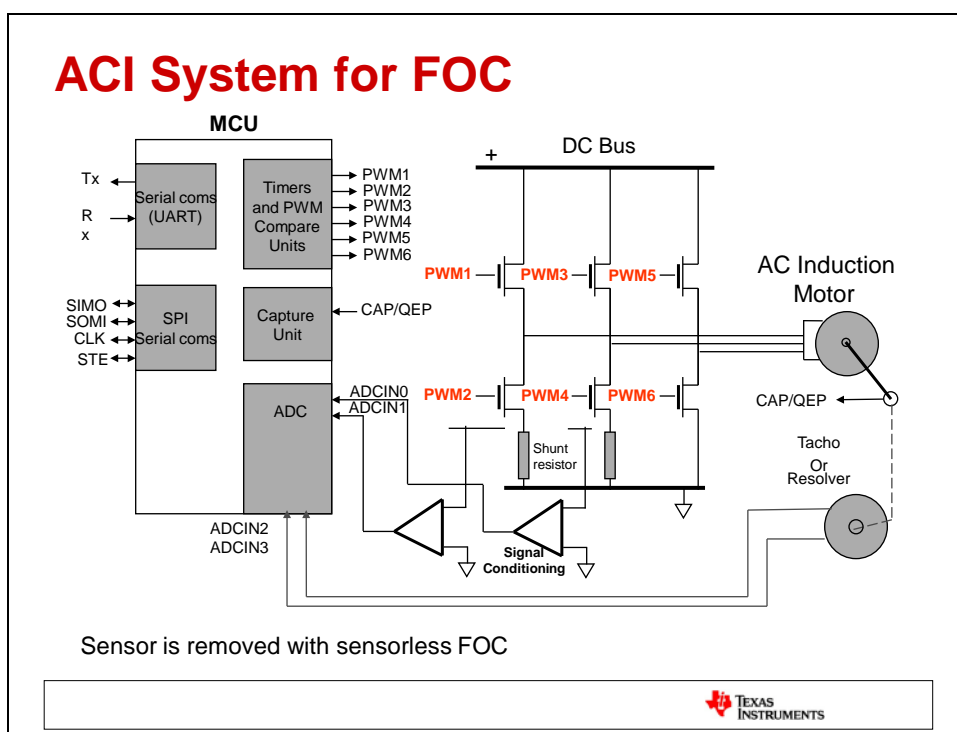
- **15% boost in torque**
- **30% less switching losses (higher efficiency)**
- **30% reduced EMI due to fewer transistor commutations**
- **Reduced harmonic copper losses**
- **Capacitor reduction**
  - The use of smaller DC link capacitor will introduce DC bus ripple
  - Controllers with processing overhead can digitally compensate for the ripple on the DC bus, allowing for a greater ripple limit
  - This smaller capacitor size can reduce system cost



# FOC Control Schematics

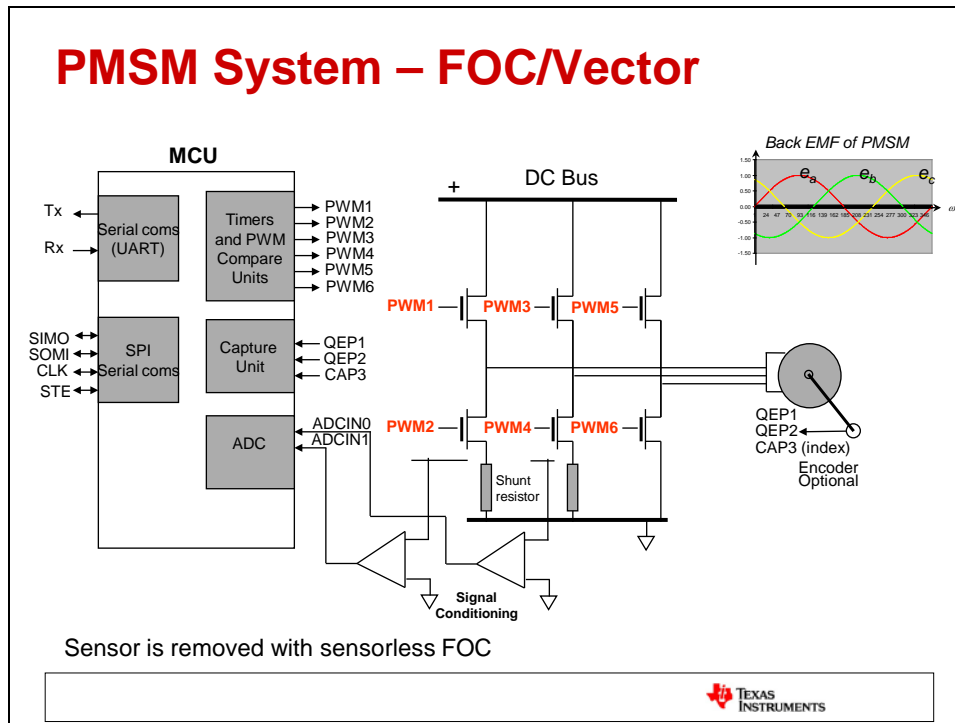
## Field Oriented ACI control

The overall system for implementation of the 3-phase ACI control unit is shown in the next slide. The ACI motor is driven by the conventional voltage-source inverter. The F2833x is generating six pulse width modulation (PWM) signals by means of space vector PWM technique for six power-switching devices in the inverter. In a “sensored” measurement mode, a tachometer or resolver is used to feedback speed and position. By contrast, in a “sensorless” measurement mode, two input currents of the PMSM ( $i_a$  and  $i_b$ ) are measured from the inverter and they are sent to the F2833x via two analog-to-digital converters (ADCs).





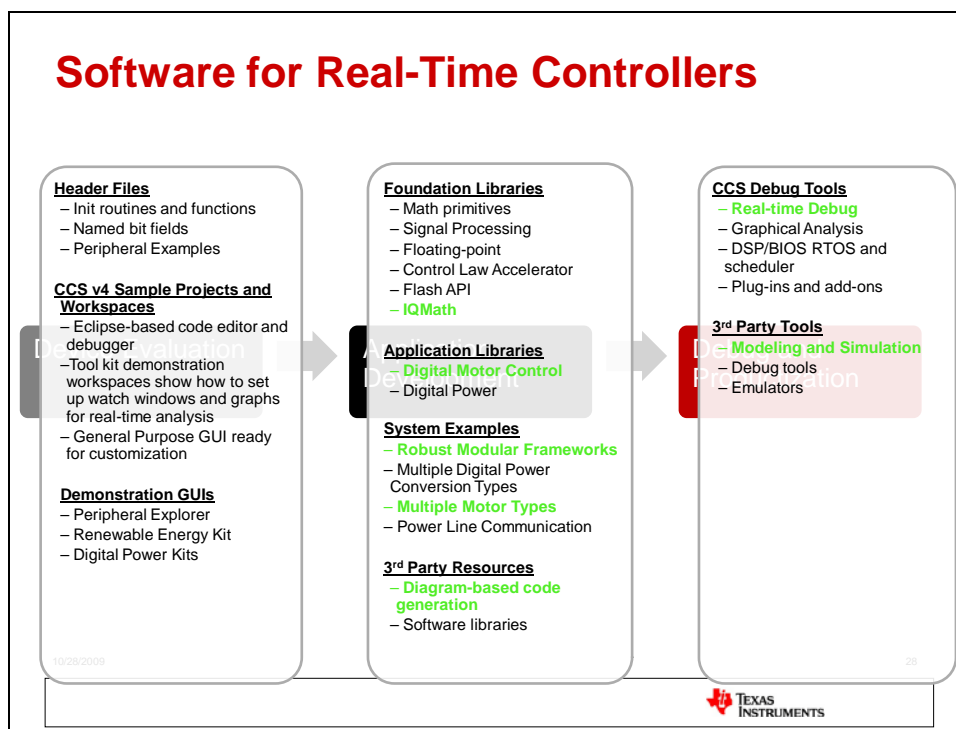
## Field Oriented PMSM control





# F2833x Features for Motor Control

## Software



## C2000 Software for DMC Concept

### Methodology

- Highest precision & most numerically accurate
- Modular libraries (C source) for easiest re-use and customization
- Removal of fixed point scaling and saturation burden
- Easiest to tune for your custom motor
- Documentation: DMC theory, software, BOM, schematics

**Customers can self serve and achieve high productivity!**

### Achieved by

- IQMath
- Application Frameworks: DMC Library & Incremental Build
- Partner Tools for simulation, GUI programming, and auto code gen

# IQMath

- **Library and Compiler Intrinsic**
  - Move your decimal point to where you need it
  - Write in floating point, compiler does all the work
- **Start-up, tuning, and debug effort are reduced**
  - Change numerical range on the fly, global or local
  - Tune for best resolution and dynamic range
  - Remove quantization effects
  - Scaling and saturation are a thing of the past
  - Better integration with simulation and code gen tools
  - Single source set to move between fixed and floating point processors
  - Easy re-use and re-tuning for new systems



## IQmath: Choose your decimal Range or Resolution?

### Based On The Required Dynamic Range Or Resolution

31 0  
SIIIIIIIIIIIIIIIIIIII.QQQQQQQQQQQQQQQQQQQ (Q15)

GLOBAL_Q	Max Val	Min Val	Resolution
28	7.999 999 996	-8.000 000 000	0.000 000 004
24	127.999 999 94	-128.000 000 00	0.000 000 06
20	2047.999 999	-2048.000 000	0.000 001

**The user selects a “Global Q” value for the entire application:**

```
#define GLOBAL_Q 24 // set in "IQmathLib.h" file
```

```
_iq Y, M, X, B;
```

```
Y = _IQmpy(M,X) + B;    // all values are in l8Q24
```

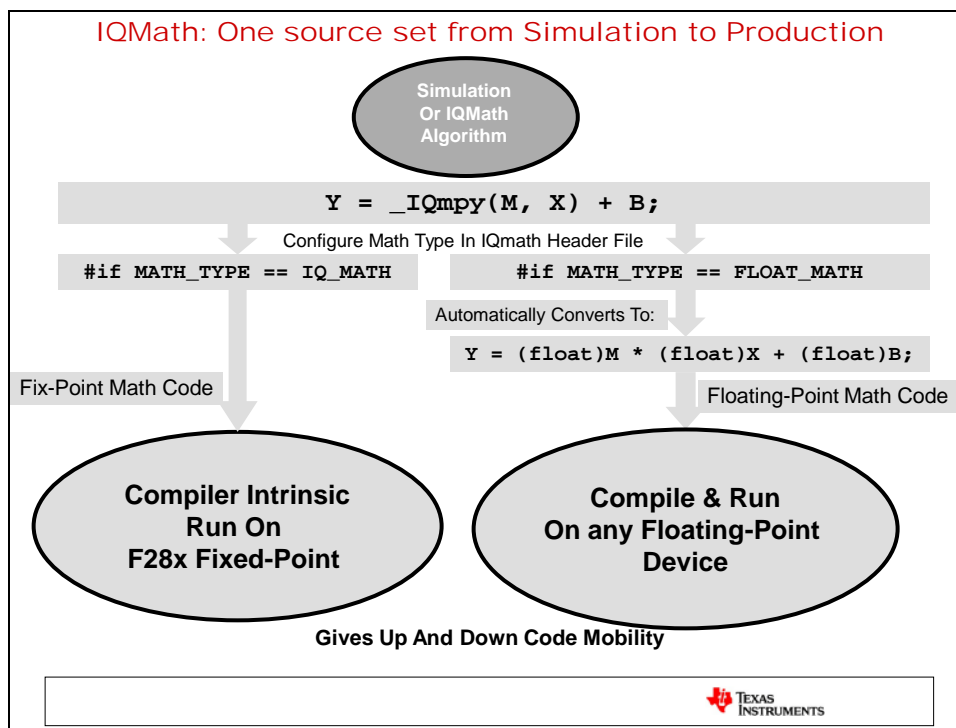
**The user can also explicitly specify the IQ value to use:**

\_iq20 Y, M, X, B;

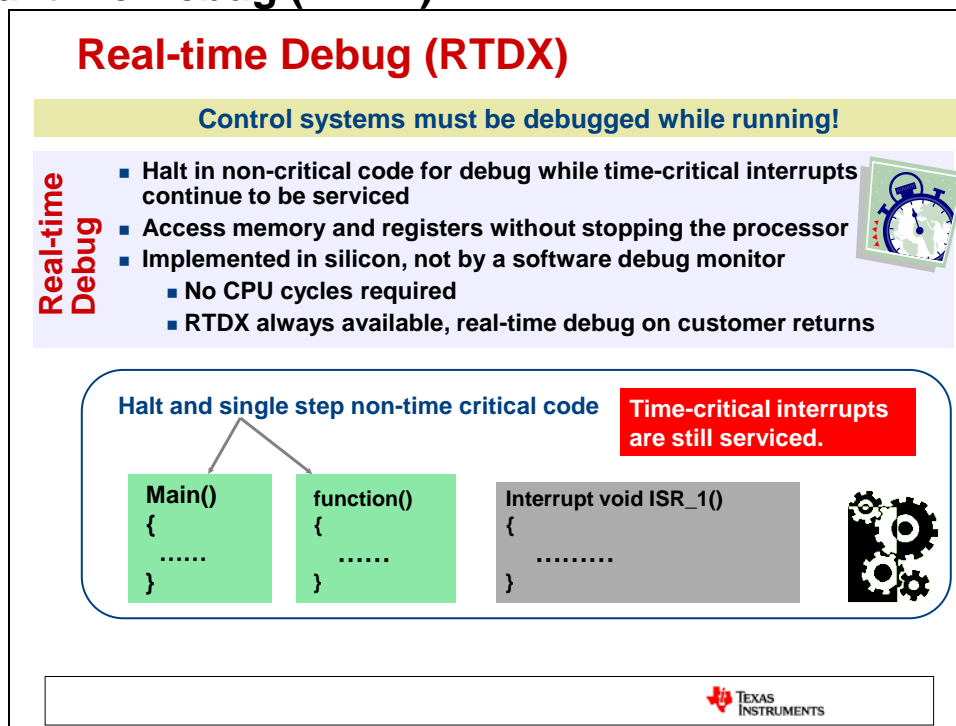
```
Y = _IQ20mpy(M,X) + B; // all values are in I12Q20
```



Probably one of the most important advantages of programming in IQMaths is the ability to switch from a fixed-point environment to the floating-point processor world. When the programmer uses a conditional compilation technique, based on “#if – else – end if” directives, the same C code can be used for fixed-point and floating-point translation.



## Real-time Debug (RTDX)



# Texas Instruments Digital Motor Control Library

## Software Modules

Texas Instruments Digital Motor Control (DMC) Library is available free of charge and can be downloaded from the Texas Instruments website. It consists of a number of useful functions for motor control applications. Among those functions, there are pure motor control modules (Park and Clark transforms, Space Vector PWM etc) as well as traditional control modules (PID controller, ramp generator etc) and peripherals drivers (for PWM, ADC and others).

Based on this DMC library, Texas Instruments has developed a number of application notes for different types of electrical motors. All applications examples are specially designed for the C2000 platform and come with a working example of the corresponding software, background information and documentation.

One branch of this library is dedicated to the F2833x and takes advantage of the 32-bit IQ-Math data format.

The following slide shows the software modules available for the C2000 family:

## DMC Library


**Name**

- aci.pdf
- aci\_fe.pdf
- aci\_se.pdf
- adc04b\_drv.pdf
- adc04u\_drv.pdf
- bdc\_pwm\_drv.pdf
- bldc\_3pwm\_drv.pdf
- cap\_event\_drv.pdf
- clarke.pdf
- comtn\_trig.pdf
- cur\_mod.pdf
- data\_log.pdf
- en\_drive\_drv.pdf
- fc\_pwm\_drv.pdf
- hall3\_drv.pdf
- i\_park.pdf
- ileg2\_dcdrv.pdf
- ileg2drv.pdf
- impulse.pdf
- mod6\_cnt.pdf
- park.pdf
- phase\_voltage\_calc.pdf
- pid\_reg3.pdf
- pwm\_dac\_drv.pdf
- pwm\_res\_drv.pdf
- qep\_no\_index\_drv.pdf
- qep\_theta\_drv.pdf
- ramp\_gen.pdf
- resolver.pdf
- rmp2cntl.pdf
- rmp3cntl.pdf
- rmp\_cntl.pdf
- smopos.pdf
- speed\_est.pdf
- speed\_frq.pdf
- speed\_prd.pdf
- svgen\_dq.pdf
- svgen\_mf.pdf
- v\_hz\_profile.pdf

**Blocks are Modular C functions**

- Variables as Inputs, Variables as Outputs
- Library of Source Code
- Most are IQ based, tune to your stability needs!

Multi-page Documentation & Theory of Operation for each module



Item	Name	Description	Format	Range(Hex)
Inputs	Ds	Direct axis(D) component of transformed signal in rotating reference frame	GLOBAL_Q	00000000-7FFFFFFF
	Qs	Quadrature axis(Q) component of transformed signal in rotating reference frame	GLOBAL_Q	00000000-7FFFFFFF
	Angle	Phase angle between stationary and rotating frame	GLOBAL_Q	00000000-7FFFFFFF (0 - 360 degree)
Outputs	Alpha	Direct axis(D) component of the transformed signal	GLOBAL_Q	00000000-7FFFFFFF
	Beta	Quadrature axis(Q) component of the transformed signal	GLOBAL_Q	00000000-7FFFFFFF

GLOBAL\_Q valued between 1 and 30 is defined in the IQmathLib.h header file.

**Other Libraries Available**

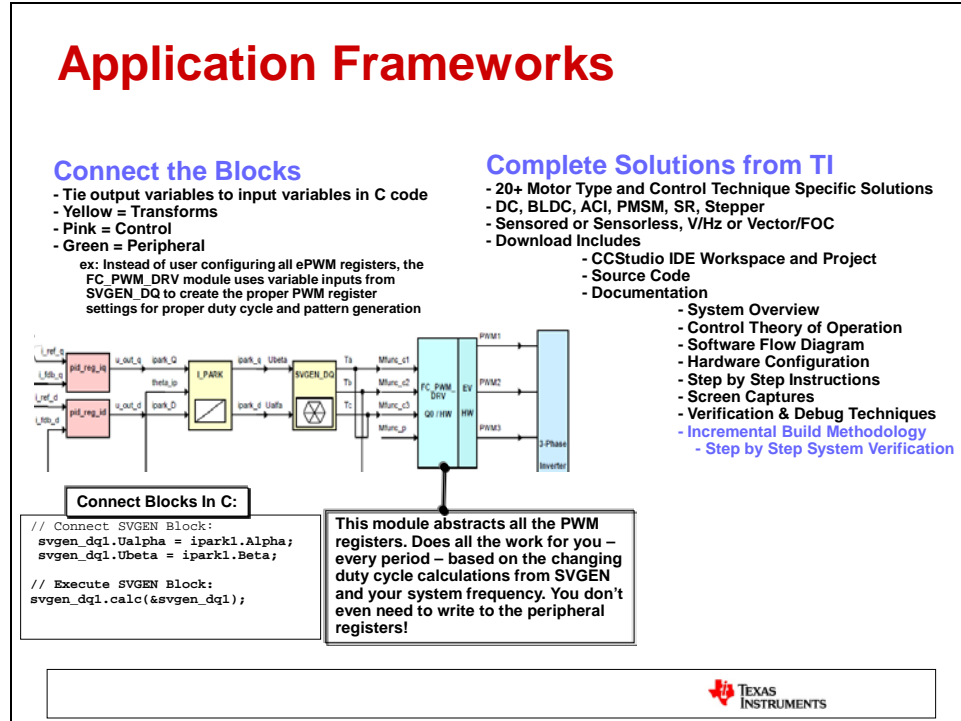
- FFT (32-bit Complex & Real)
- Filters (FIR, IIR)
- QMath (Trig, SQRT, INV, LOG, DIV)
- IQMath Virtual Floating Point (Conversion, Arithmetic, Trig, Math)
- Signal Generation (Sine, Ramp, Trapezoidal)
- Digital Power

18 - 20

F28333x - Digital Motor Control

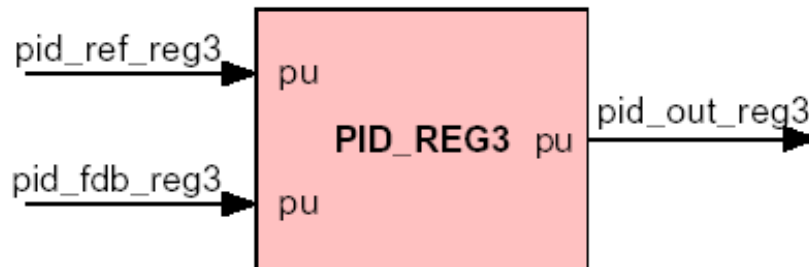
## The Application Framework

All Digital Motor Control Library solutions are based on a framework system, shown in the following slide. Although the modules are written in optimized IQMath code, all of them can be accessed using a C language interface.



What the user has to do is simply to select the correct blocks, to define the variables for input and output lines of the corresponding blocks and to connect these “lines” by passing variables. All modules are supplied with a dedicated documentation file.

For example, the file “pid\_reg3.pdf” explains the interface and the background of the PID-controller:



All functions are coded for 32-bit variables in IQ-Math-format. The functions are used as instances of a predefined object class, based on a structure definition in a header file.

## Texas Instruments DMC Solutions

Texas Instruments offers a set of more than 20 complete solutions for different types of motors, switching and control techniques, all based on this application framework.

### Application Frameworks, Hardware, and Application Notes

C2000			
MOTOR TYPE	CONTROL	FEEDBACK	SOLUTION
STEPPER	Microstepping	Sensorless	SPRAAU7
DC	Speed & Position	Sensored	SPRC177, SPRC214
ACI	V/F	Sensored	SPRC130, SPRC194
	FOC	Sensored	SPRC077, SPRC207
	FOC	Sensorless	SPRC078, SPRC195, SPRC922
BLDC	Trapezoidal	Sensored	SPRC175, SPRC213
	Trapezoidal	Sensorless	SPRC176, SPRC196
PMSM	V/F	Sensored	SPRC129, SPRC210
	FOC - Resolver	Sensored	SPRC178, SPRC211
	FOC	Sensored	SPRC179, SPRC212
	FOC	Sensorless	SPRC128, SPRC197, SPRC922, TMD51MTRPFCKIT, TMD52MTRPFCKIT
SWITCHED RELUCTANCE	Two Quadrant	Sensorless	SPRA600
OTHER			DMC Library: SPRC080, SPRC215 Designing High Performance DMC: SPRT528



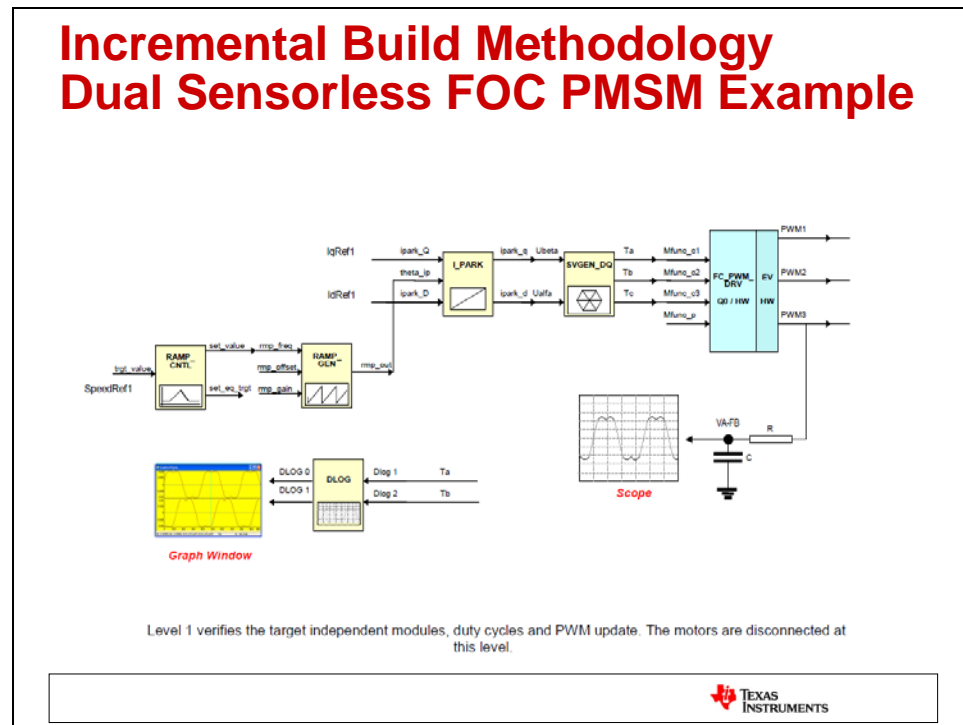
These solutions follow a simple principle for testing the software, accessing the power drives and closing the control loop: an incremental build methodology. The basic idea is to define a macro and to use a conditional compilation (called: “Build Level”) to include more and more modules into the final machine code. Such a technique is very helpful when the user tests a motor drive system for the very first time.

The following slides explain this sequential method with the example of a PMSM Field Oriented Control System.

# Example: PMSM Framework

## Build Level 1

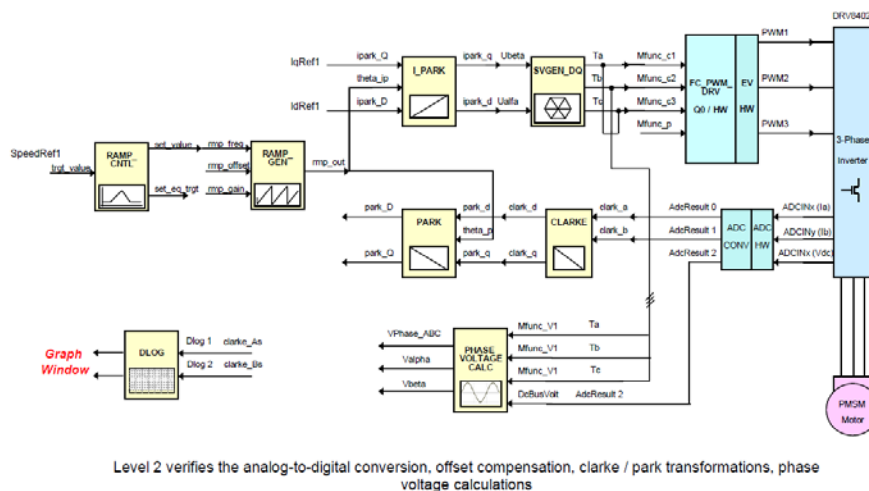
Build Level 1 is used to verify the target independent modules, such as PWM frequency, duty cycles and updates of the PWM unit. The motors are disconnected at this level. Two software modules “RAMP\_GEN” and “RAMP\_CNTL” are used to stimulate the PWM system via the inverse PARK module and the Space Vector Generator module.



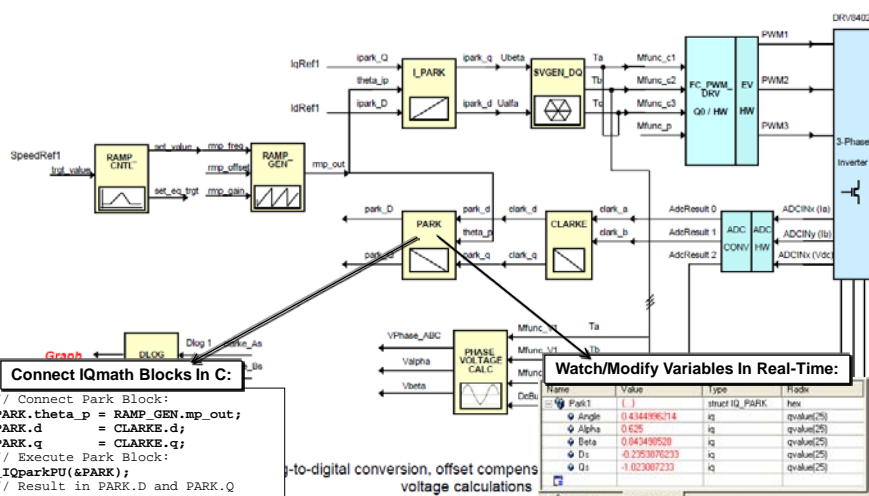
An oscilloscope is used to monitor the shape of the PWM signals and the pulses series generated by the SVGEN module.

## Build Level 2

### Incremental Build Methodology Dual Sensorless FOC PMSM Example



### Incremental Build Methodology Dual Sensorless FOC PMSM Example











## C2000 Motor Control Hardware

To start an exploration for motor control, based on C2000 controllers, Texas Instruments offers a set of low cost tools:

### controlSTICK: Low Cost Evaluation

#### \$39 Kit includes

- Simple USB memory stick form factor evaluation tool
  - Piccolo F28027
  - Onboard USB JTAG emulation
  - Header pins provides access to most Piccolo pins
- 11 example projects explain most Piccolo peripherals
- Jumpers and patch cords to easily connect pins together
- USB extension cable
- Code Composer Studio V3.3 with 32KB code size limit
- Complete hardware documentation
  - Gerbers, schematics, etc

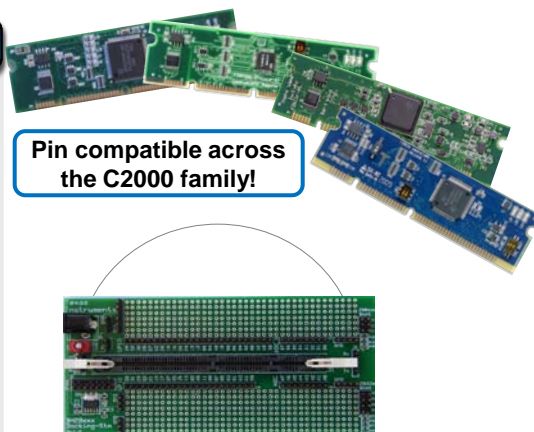


### controlCARD: Modular, Robust, Standard

#### controlCARD












- Low cost, small form factor
- Standard DIMM interface
  - Includes analog I/O, digital I/O, and JTAG signals available at DIMM interface
- Robust design
  - Noise filter at ADC input pins
  - Ground plane
  - Isolated UART communication
  - Supply pin decoupling
- All life support circuitry
  - Clock, Power Supply, LDO, etc
- Multiple versions available
  - Piccolo F28027
  - Piccolo F28035
  - Delfino F28335
  - Delfino C28436
  - F2808
  - F28044
  - \$49-69


Pin compatible across the C2000 family!



A set of base boards allows the user to go deeper into different application areas, such as Digital Power Supply or Digital Motor Control.

## controlKIT: controlCARD + base board

Device Evaluation	Application Development
 <p><b>Experimenter's Kit - Flash</b> TMDXDCK28027 TMDXDCK28035 TMDXDCK2808 TMDXDCK28335 \$79-\$159</p>	 <p><b>Digital Power Experimenter's Kit</b> TMDSDCDC2KIT \$229</p>
 <p><b>Experimenter's Kit - RAM</b> TMDXDCK28343 TMDXDCK28346-168 \$189</p>	 <p><b>Digital Power Developer's Kit</b> TMDSDCDC8KIT \$325</p>
 <p><b>Peripheral Explorer</b> TMDSPREX28335 \$179</p>	 <p><b>AC/DC Developer's Kit</b> TMDSACDCKIT \$695</p>
 <p><b>controlKITs Include</b></p> <ul style="list-style-type: none"> <li>• controlCARD + Base Board</li> <li>• CCStudio IDE v3.3 32KB code size limit</li> <li>• Example Software with lab document</li> <li>• Power Supply and Cables</li> </ul>	 <p><b>Resonant DC/DC Developer's Kit</b> TMDSRESCKIT \$229</p>
 <p><b>Developer's Package</b></p> <ul style="list-style-type: none"> <li>• Schematics (source and .PDF files)</li> <li>• Bill of materials (BOM)</li> <li>• Gerber files to freely use or modify</li> <li>• Pin-out table showing all key signals</li> <li>• DIMM100 pin/socket mechanical details</li> </ul>	 <p><b>Renewable Energy Developer's Kit</b> TMDSENRYKIT \$349</p>
	 <p><b>Sensorless FOC DMC + PFC Developer's Kit</b> TMDS1MTRPFCKIT 1 Motor TMDS2MTRPFCKIT 2 Motor \$369/\$399</p>




For Digital Motor Control the following package includes all you need to experiment with PMSM motors.

## Sensorless FOC and PFC Developer's Kit


**\$399 Motor Control and PFC Kit Includes**

- Piccolo F28035 controlCARD
- Sensorless Sinusoidal SVPWM based Field Oriented Control
- Single or Dual Axis Operation
- Integrated Digital Power Factor Correction
- Hardware Features
  - 100W 2 phase interleaved power factor correction stage
  - 2 x 60 W motor driver stages based on TI DRV8402 motor driver chips
  - On board isolated XDS100 JTAG emulation
- Software Lab Projects
  - Standalone PFC
  - Dual Axis
  - Single Axis + PFC
  - Dual Axis + PFC (coming soon)

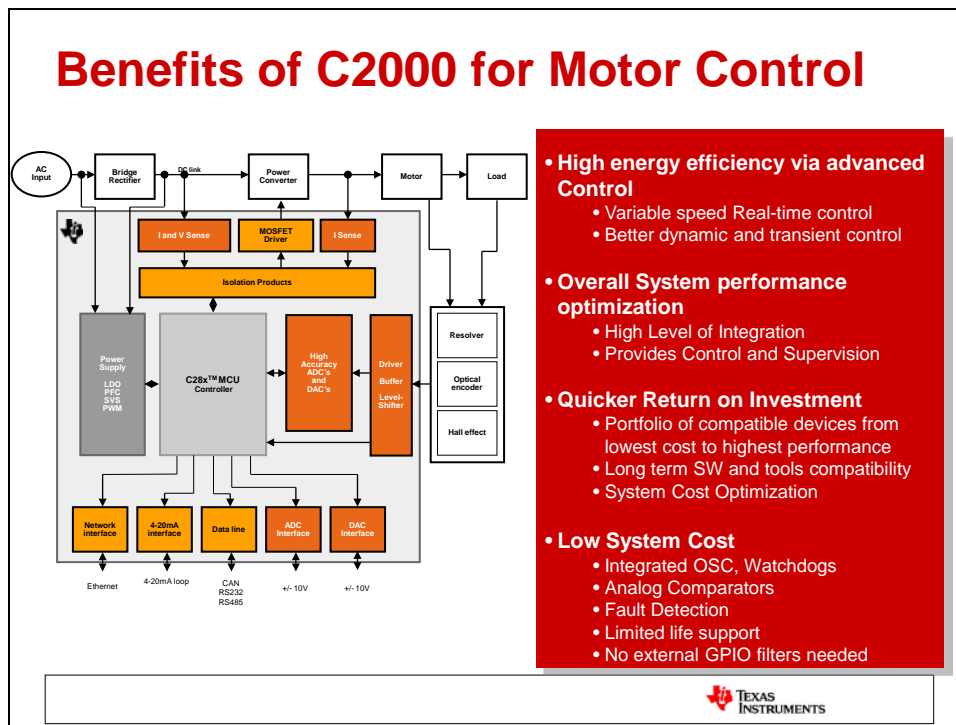


**No external emulator required!**

TMDS2MTRPFCKIT(2 motors)  
TMDS1MTRPFCKIT(1 motor)



## Summary



## Motor Control Development Kit

The Digital Motor Control Kit “TMDS2MTRPFCKIT” is an ideal target to experiment with control loops for electrical motors.

This kit comes with a set of documentation, including all the software that is needed to build software projects as described in this Chapter. You can also download this software from the Texas Instruments website ([www.ti.com](http://www.ti.com)):

- Search for literature number “sprc922.zip” to obtain the board specific software
- Search for literature number “sprc675.zip” to get the software baseline for this kit
- Search for literature number “SPRUGQ1” to get the Quick Start Guide for the board.