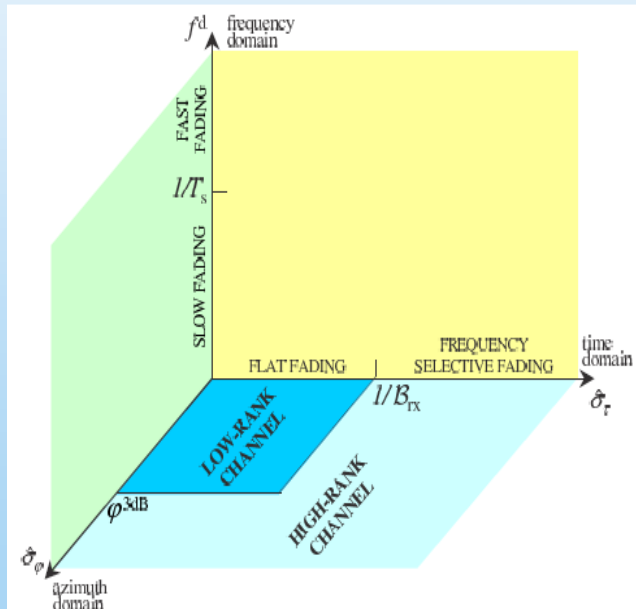
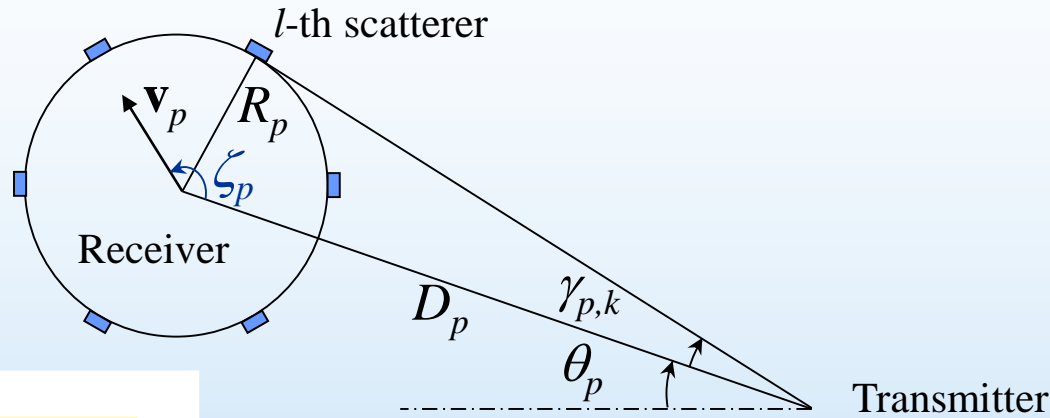




COMPUTER ENGINEERING



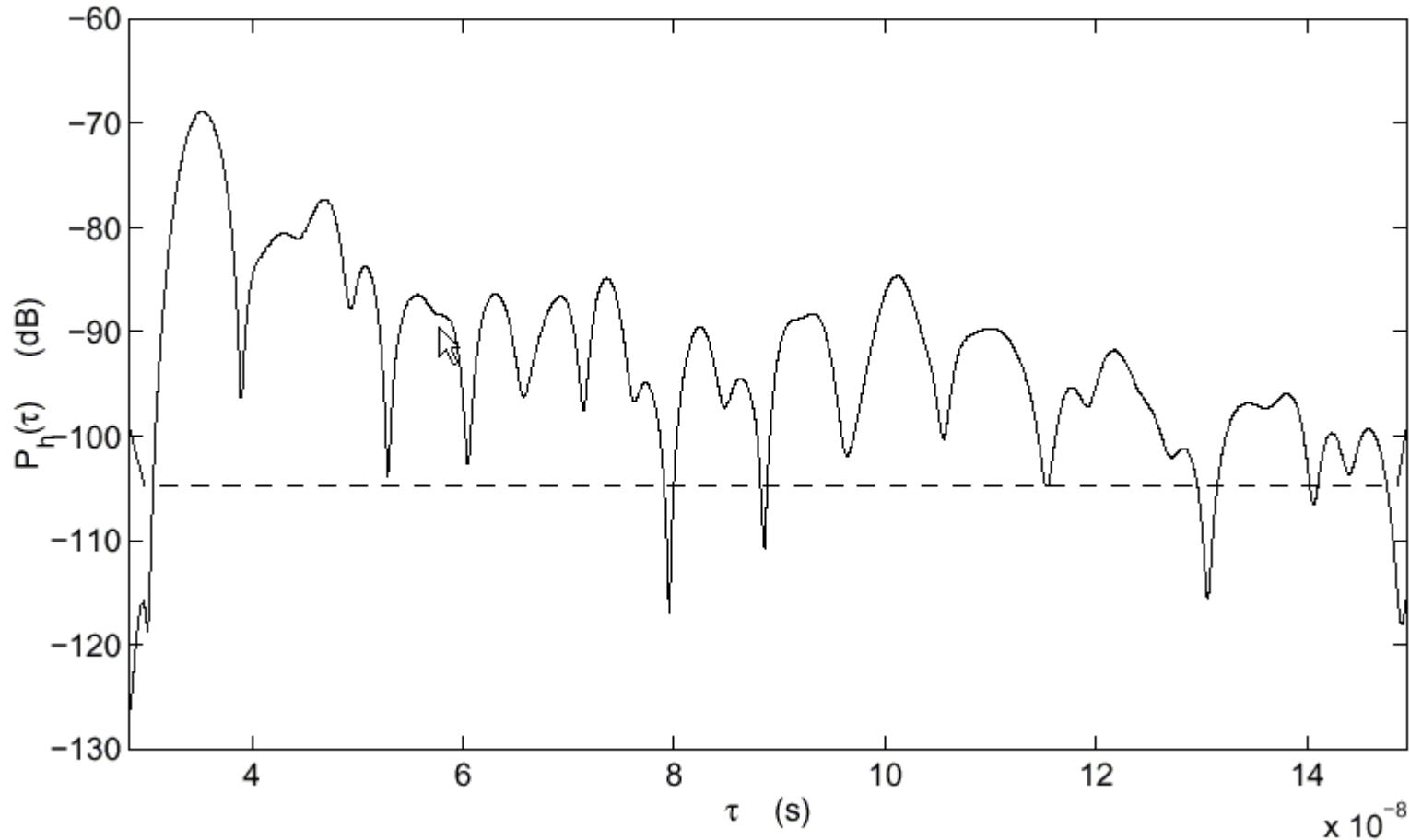
# The Mobile Radio Channel



$$h(t, \tau) = \sum_{p=1}^P \sum_{l=1}^{L_p} \alpha_{l,p}(t) e^{j\phi_{l,p}(t)} \delta(\tau - \tau_{l,p})$$

# Power Delay Profile

- Assume to transmit an ideal pulse at  $t=0$ . Measured power response.



- Average delay: 
$$\mu_\tau = \frac{\int_0^\infty \tau \phi_c(\tau) d\tau}{\int_0^\infty \phi_c(\tau) d\tau}$$

- Delay spread: 
$$\sigma_\tau = \sqrt{\frac{\int_0^\infty (\tau - \mu_\tau)^2 \phi_c(\tau) d\tau}{\int_0^\infty \phi_c(\tau) d\tau}}$$

- Be  $\Phi_T(\Delta f)$  function measures the frequency correlation of the channel. We define the coherence band,  $B_c$ , the frequency separation,  $\Delta f$ , for which  $\Phi_T(\Delta f)$  assumes a prefixed value (usually 0.5). If the signal bandwidth,  $W$ , does not exceed  $B_c$ , the linear distortion due to the channel is not significant.
- The coherence band is proportional to the reciprocal of the delay spread:

$$B_c \equiv \frac{1}{\sigma_\tau}$$

- Be  $\Phi_H(\nu)$  the functions that provides the behavior of the output power as a function of the Doppler frequency,  $\nu$ .
- The bandwidth,  $B_d$ , within which this function assumes significant values is called **Doppler spread**.
- The inverse of this bandwidth is called **coherence time** (time interval in which there is a significant correlation between the characteristics of the channel).
- Observe that  $B_d$  is proportional to the mobile speed,  $\nu$ ,  $B_d \equiv \nu/\lambda = \nu f_c/c$ , being  $f_c$  the carrier frequency and  $c$  the speed light in the vacuum.

- Suppose to model the channel with  $N$  paths, each of which has an attenuation  $\alpha_i$  and a delay  $\tau_i$ . The time varying channel pulse response is:

$$c(t, \tau) = \sum_{i=1}^N \alpha_i(t) e^{-j2\pi f \tau_i(t)} \delta(\tau - \tau_i(t))$$

- Weight of the generic path:

$$p_i = \frac{\alpha_i^2}{\sum_{n=1}^N \alpha_n^2}$$

- Average delay and delay spread:

$$\mu_\tau = \sum_{i=1}^N p_i \tau_i \qquad \sigma_\tau = \sqrt{\sum_{i=1}^N p_i (\tau_i - \mu_\tau)^2}$$

- Signal bandwidth less than coherence bandwidth.
- Non-selective frequency fading.
- Received signal:

$$r(t) = \alpha e^{j\phi} u_m(t) + z(t), 0 \leq t \leq T$$

- where  $\alpha$  e  $\phi$  are random variables, while  $z(t)$  represents the Gaussian noise.

- Hypotheses:
  - Incident field obtained as a result of many components coming from all directions.
- Clarke model:
  - The real and imaginary components of the field are random variables Gaussian (average zero if the transmitter and receiver are not in visibility - Rayleigh fading, not zero otherwise – Rice fading), uncorrelated.
  - The autocorrelation coefficient (normalized autocovariance) of each of the two components is expressed by:  $\rho(\tau) = J_0(2\pi f_d \tau)$   
( $J_0$  it is the Bessel's function of first kind and of order 0, and  $f_d$  is the Doppler spread).



- The **autocorrelation coefficient** (normalized autocovariance) of amplitude and power are both given by the following expression (for the amplitude it is an approximate description, while for power it is an exact expression):

$$\rho(\tau) = J_0^2(2\pi f_d \tau)$$

- The **power spectrum** is given by:

$$S(f) = \frac{P}{64\pi} \frac{1}{f_d} \mathbf{K} \left( \sqrt{1 - \left( \frac{f}{2f_d} \right)^2} \right), \quad 0 \leq |f| \leq 2f_d$$

where  $\mathbf{K}(x)$  is the complete elliptic integral of the first type.

- In the hypothesis of non selective channel (signal bandwidth less than the coherence bandwidth), the signal undergoes a complex attenuation,  $\alpha(t)$ , variable over time (fading).
- Observing the autocorrelation coefficient, we note that fading values spaced by  $\tau \ll 0.1/f_d$  are strongly correlated, while values spaced by  $\tau \gg 0.1/f_d$  are uncorrelated.
- Defined  $T$  the time interval of interest (for example duration of a symbol or a package), fading is said to be slow if  $T \ll 0.1/f_d$  (consecutive symbols undergo the same attenuation), and fast if  $T \gg 0.1/f_d$  (consecutive symbols undergo independent attenuations).

- The amplitude of the received signal,  $\rho$ , is a random variable, the probability density function (pdf) of which is given by (Rice):

$$f_{\rho}(\rho|P, K) = (1 + K)e^{-K} \frac{\rho}{P} e^{-\frac{1+K}{2P}\rho^2} I_0\left(\rho \sqrt{\frac{2K(1+K)}{P}}\right)$$

where  $K$  is the **Rice factor**, (ratio between the power of the direct path and the power received through reflections),  $P$  is the average received power, and  $I_0$  is the modified Bessel function of first kind and order 0.

- The received,  $p = \rho^2/2$ , is a random variable the pdf of which is a non central  $\chi^2$  distribution with 2 degrees of freedom::

$$f_p(p|P, K) = (1 + K) \frac{e^{-K}}{P} e^{-\frac{1+K}{P}p} I_0\left(\sqrt{4K(1+K)\frac{p}{P}}\right) \quad \sigma^2 = P^2 \frac{1+2K}{(K+1)^2}$$

- When  $K=0$  (no direct path) we obtain the Rayleigh fading for which:

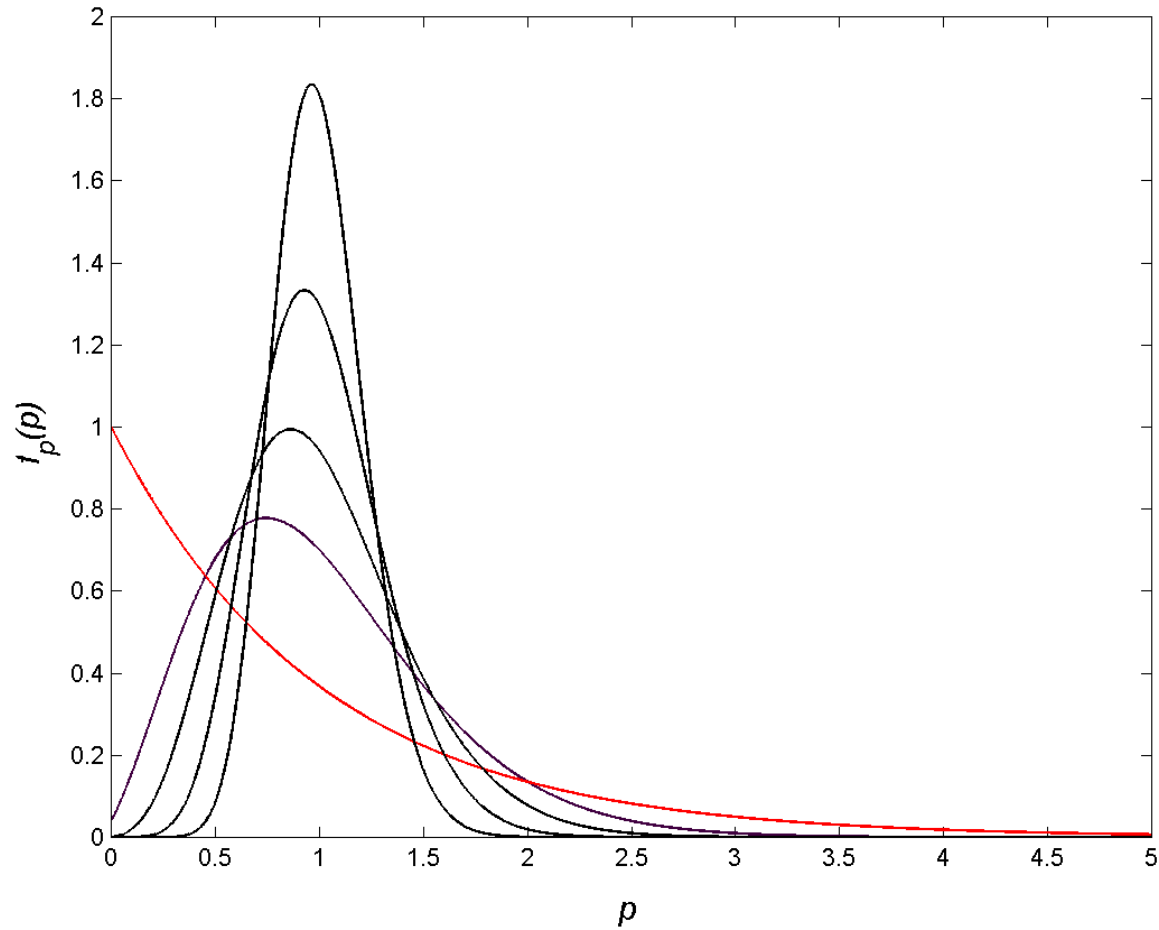
$$f_\rho(\rho|P) = \frac{\rho}{P} e^{-\frac{\rho^2}{2P}} \quad f_p(p|P) = \frac{1}{P} e^{-\frac{p}{P}}$$

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Power pdf for  $P=1$  and  $K=0,5,10,20,40$  .

- The attenuation due to obstacles, foliage follow a log-normal type statistic (shadowing).
- Given a random variable  $V \geq 0$ , it follows a log-normal statistic if the variable  $U = \ln(V)$  (or  $U = 10 \log(V)$ , or  $U = 20 \log(V)$ ) is a Gaussian variable.

- Average received power in dBm

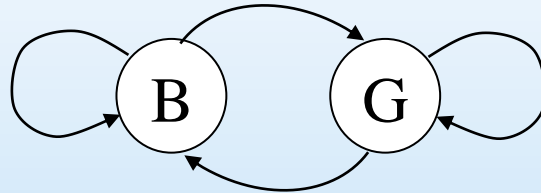
$$U = 10 \log_{10} P_R \quad \mu_{\text{dB}} = E[U] \quad \sigma_{\text{dB}}^2 = E[(U - \mu_{\text{dB}})^2]$$

$$f_U(U) = \frac{1}{\sqrt{2\pi\sigma_{\log n}^2}} e^{-\frac{(U - \mu_{\text{dB}})^2}{2\sigma_{\text{dB}}^2}} \Rightarrow f_{P_R}(P_R) = \frac{C}{\sqrt{2\pi\sigma_{\text{dB}}^2}} \frac{1}{P_R} e^{-\frac{(10 \log_{10} P_R - \mu_{\text{dB}})^2}{2\sigma_{\text{dB}}^2}} \quad \text{with } C = 10 \log_{10} e$$

defined  $\mu = \mu_{\text{dB}}/C$  and  $\sigma = \sigma_{\text{dB}}/C$ , we have  $f_{P_R}(P_R) = \frac{1}{\sqrt{2\pi\sigma^2}} \frac{1}{P_R} e^{-\frac{(\log P_R - \mu)^2}{2\sigma^2}}$

from which  $E[P_R] = \exp(\mu + \sigma^2/2)$  and  $\text{var}[P_R] = \exp(2\mu + \sigma^2)(\exp(\sigma^2) - 1)$

- Assume that the long-term variations in the received signal may be described by a semi-Markov chain including two distinct states, Good (with a limited shadowing) and Bad (with severe shadowing) [Rec. ITU-R P.681-11].



- The duration of each state is considered to be log-normally distributed.

$$f_X(x) = \frac{1}{x\sqrt{\pi\sigma_i^2}} e^{-\frac{(\log x - \mu_i)^2}{2\sigma_i^2}}$$

where:

$i = G$  for Good states

$i = B$  for Bad states

$\mu_G$  and  $\sigma_G$ : mean and standard deviation for Good state

$\mu_B$  and  $\sigma_B$ : mean and standard deviation for Bad state.



- The signal envelope in the Good and Bad states follows a Loo distribution taking into account both fading and shadowing.

$$f_{\text{Loo}}(x) = x \frac{20 \log_{10} e}{\sigma_i^2 \Sigma_{A_i} \sqrt{2\pi}} \int_0^{\infty} \frac{1}{a} \exp \left[ -\frac{(20 \log_{10} a - M_{A_i})^2}{2 \Sigma_{A_i}^2} - \frac{x^2 + a^2}{2 \sigma_i^2} \right] I_0 \left( \frac{ax}{\sigma_i^2} \right) da$$

- where
  - $M_{A_i}$ : mean of the direct signal
  - $\Sigma_{A_i}$ : standard deviation of the direct signal
  - $M_{P_i} = 10 \log_{10}(2\sigma_i^2)$  dB: mean of the reflections (multipath)
- Without shadowing,  $a=A$ , we obtain the Rice fading

$$f_{\text{Rice}}(x) = \frac{x}{\sigma_i^2} \exp \left[ -\frac{x^2 + A^2}{2\sigma_i^2} \right] I_0 \left( \frac{Ax}{\sigma_i^2} \right)$$

with  $A = \sqrt{\frac{2KP}{1+K}}$ ,  $\sigma_i^2 = \frac{P}{1+K}$ , from which  $K = \frac{A^2}{2\sigma_i^2}$   $P = 2\sigma_i^2 + A^2$

- Without the reflections ( $\sigma_i=0$ ) we obtain the lognormal density probability

$$f_{\text{lognormal}}(x) = \frac{1}{x} \frac{20 \log_{10} e}{\Sigma_{A_i} \sqrt{2\pi}} \exp \left[ -\frac{\left(20 \log_{10} x - M_{A_i}\right)^2}{2 \Sigma_{A_i}^2} \right]$$

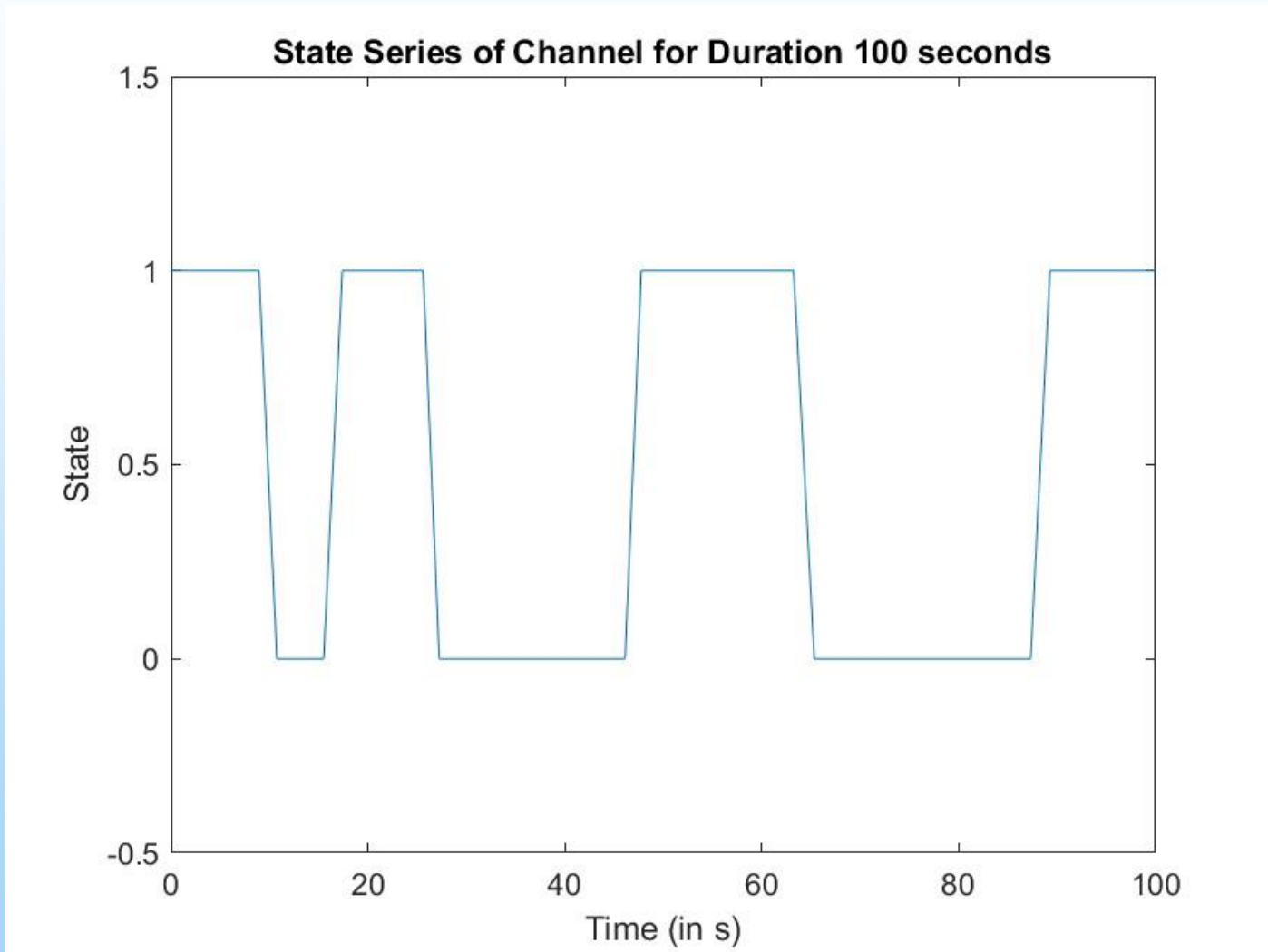
with moments

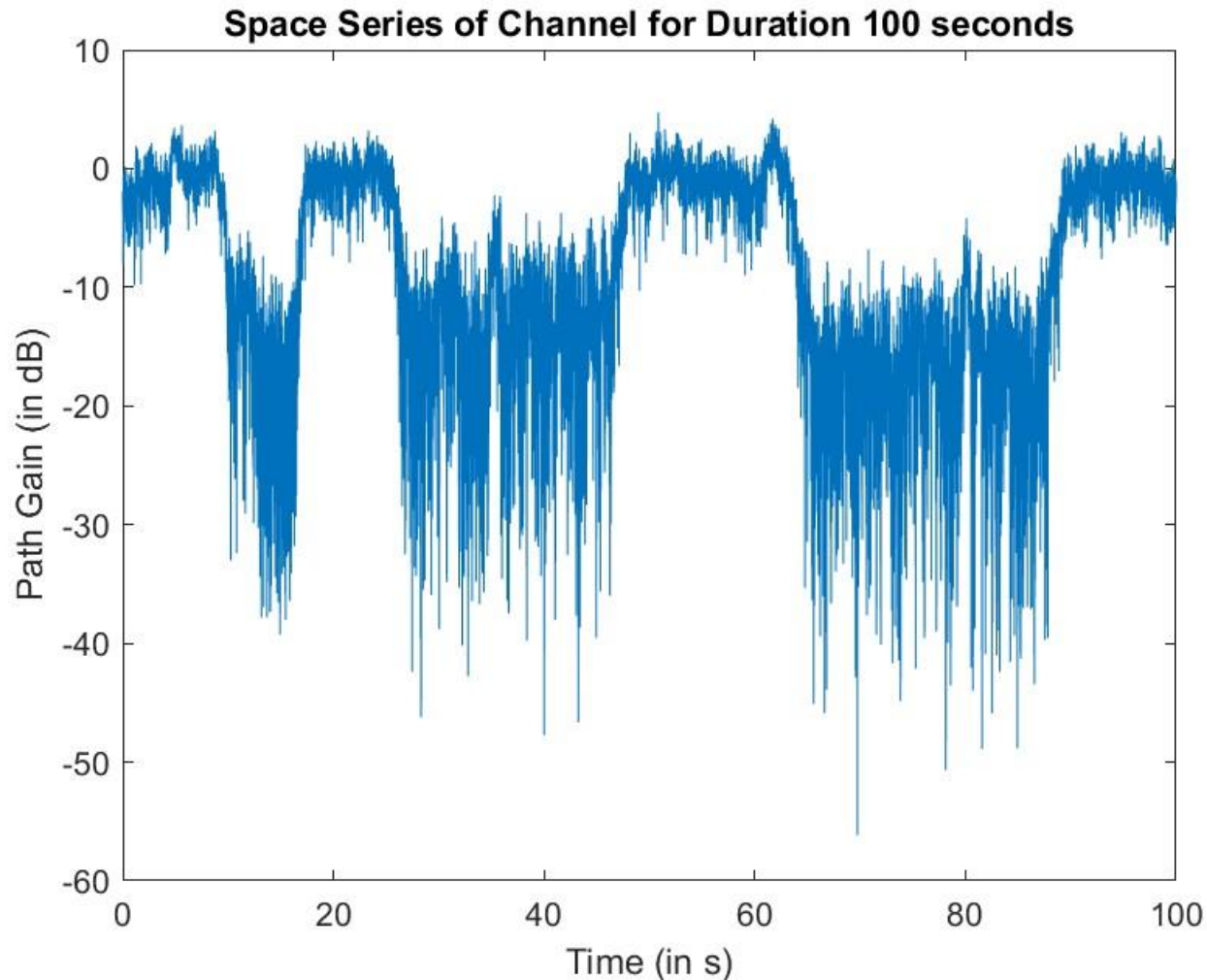
$$E[x] = 10^{M_{A_i}/20 + (\Sigma_{A_i}/20)^2 \log(10)/2}$$

$$\text{var}[x] = 10^{M_{A_i}/20 + (\Sigma_{A_i}/20)^2 \log(10)} \left( 10^{(\Sigma_{A_i}/20)^2 \log(10)} - 1 \right)$$

- The data sets provided for ITU-R P.681-11 channel are applicable for frequency range of 1.5 GHz to 20 GHz.
- Matlab release 2024a includes a channel simulator for the ITU channel.
- These parameters are requested to model a specific scenario
- Environment (values: "Urban" (default) | "Suburban" | "RuralWooded" | "Village" | "Residential" | "Highway" | "Rural" | "Train" | "Custom")
- Carrier frequency
- Elevation angle
- Speed of the ground terminal
- Azimuth orientation of the ground terminal

- Example
  - Environment type: “Urban”.
  - Carrier frequency:  $3.8e9$  Hz.
  - Elevation angle:  $45^\circ$
  - Speed of the ground terminal: 2 m/s
  - Azimuth orientation of the ground terminal:  $0^\circ$



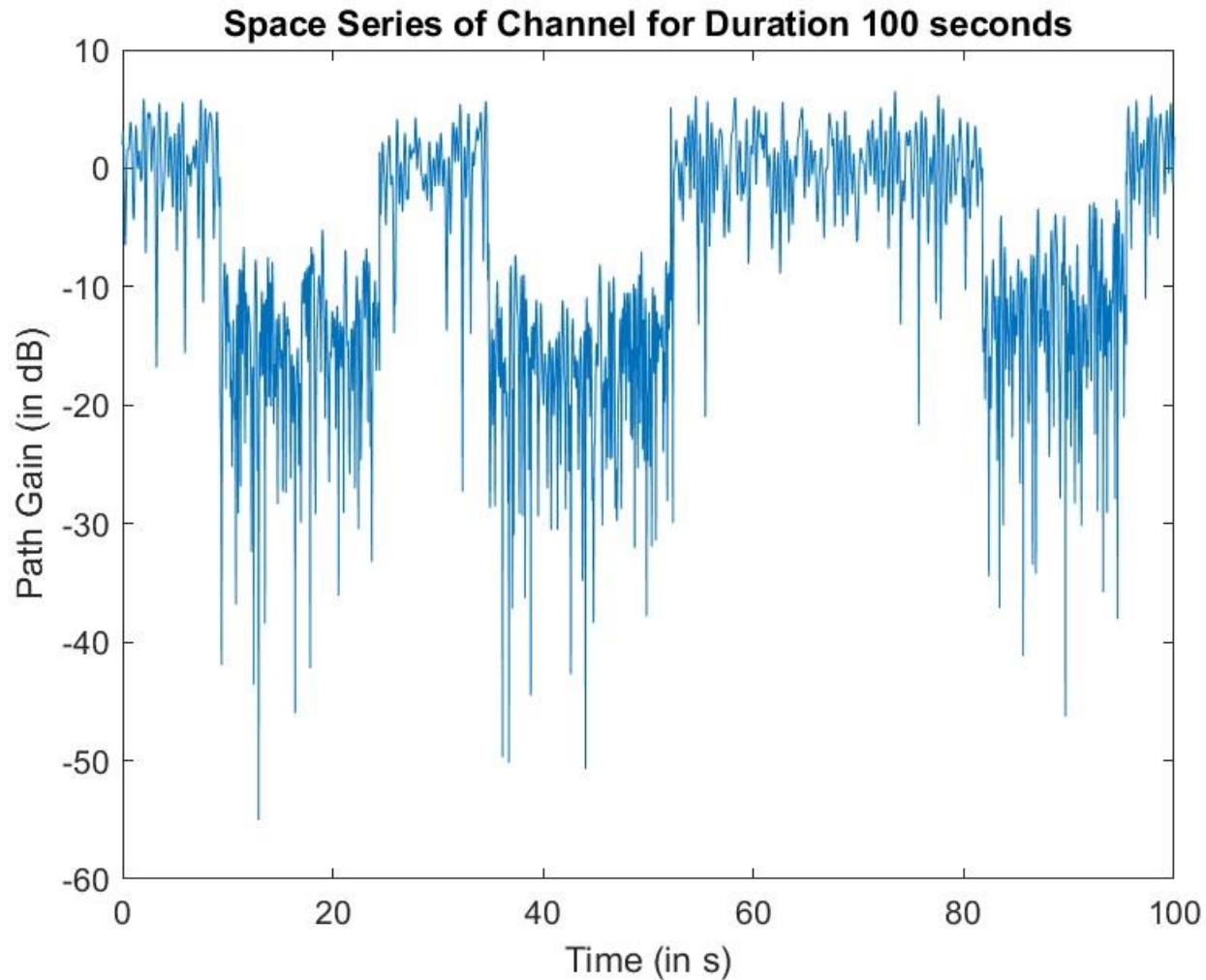


- E. Lutz, D. Cygan, M. Dippold, F. Dolainsky, and W. Papke, "The land mobile-satellite communication channel-recording, statistics, and channel model", IEEE Trans. Veh. Technol., vol 40, no. 2, pp. 375-386, 1991.
- Suitable for L band (1.54 GHz).
- Uses Rician distribution in good state and Rayleigh with log-normal distribution in bad state.
- The state duration distribution has to be specified.
- Available in Matlab with the following parameters
  - Rician K-factor
  - Lognormal fading parameters
  - State duration distribution
  - Mean state duration
  - Maximum Doppler shift

- Example
  - Rician K-factor: 5.5 dB
  - Lognormal fading parameters: [-13.6 3.8]
  - State duration distribution: "Exponential"
  - Mean state duration: [21 24.5]
  - Maximum Doppler shift: 2.8538 Hz.



## Two-state Lutz channel (3)



- Fading depth determined by Matlab simulator

